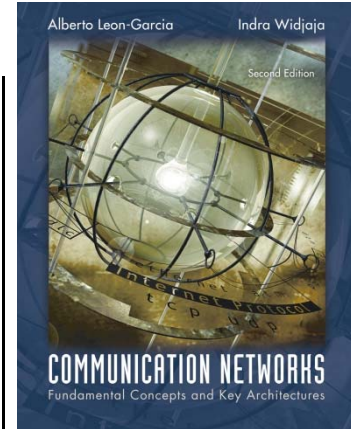


Chapter 3

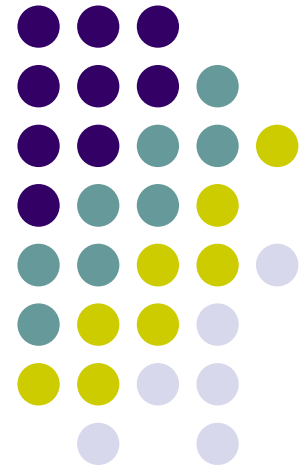
Digital Transmission Fundamentals



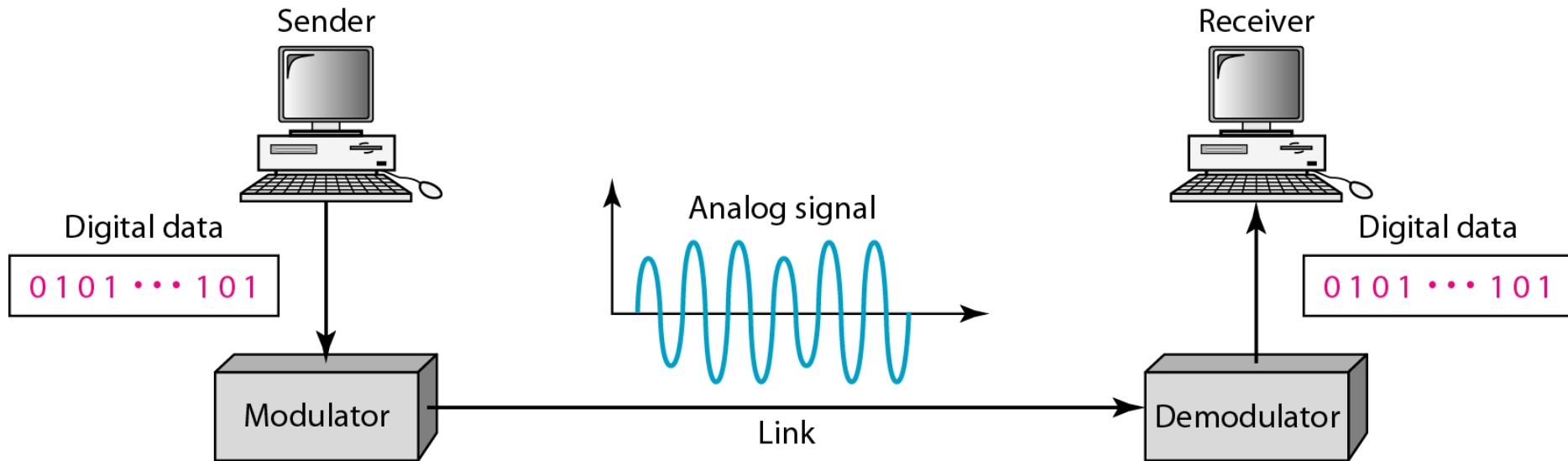
Modems and Digital Modulation

CSE 3213, Winter 2010

Instructor: Foroohar Foroozan



Modulation of Digital Data

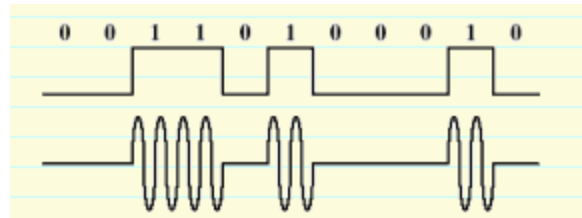


Modulation of Digital Data

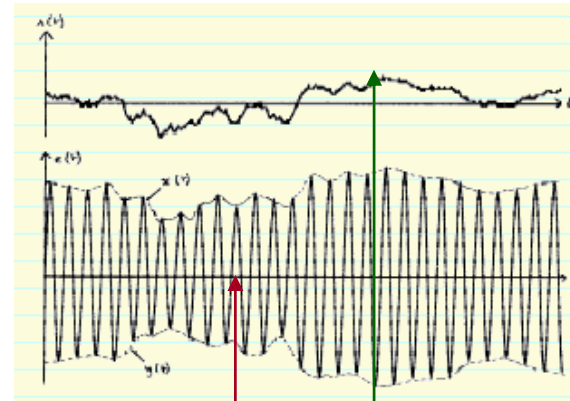


Modulation

- process of converting digital data or a low-pass analog signal to band-pass (higher-frequency) analog signal

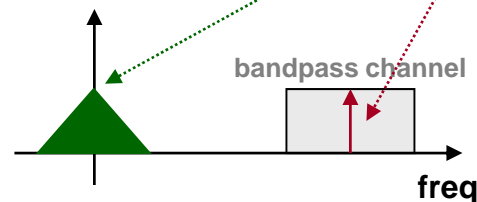


Digital-to-analog modulation.



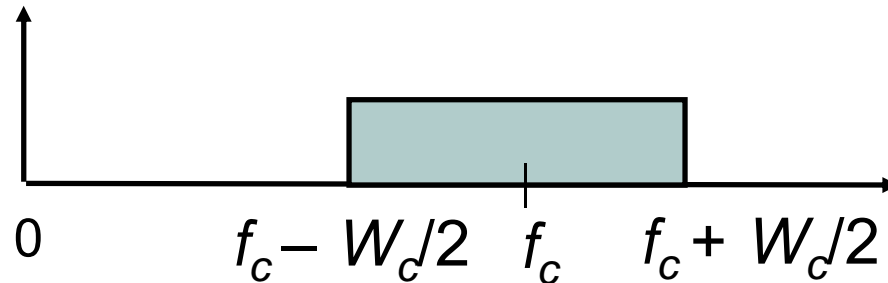
Analog-to-analog modulation.

- aka carrier frequency - modulated signal - high frequency signal that acts as a basis for the information signal
 - information signal is called modulating signal





Bandpass Channels

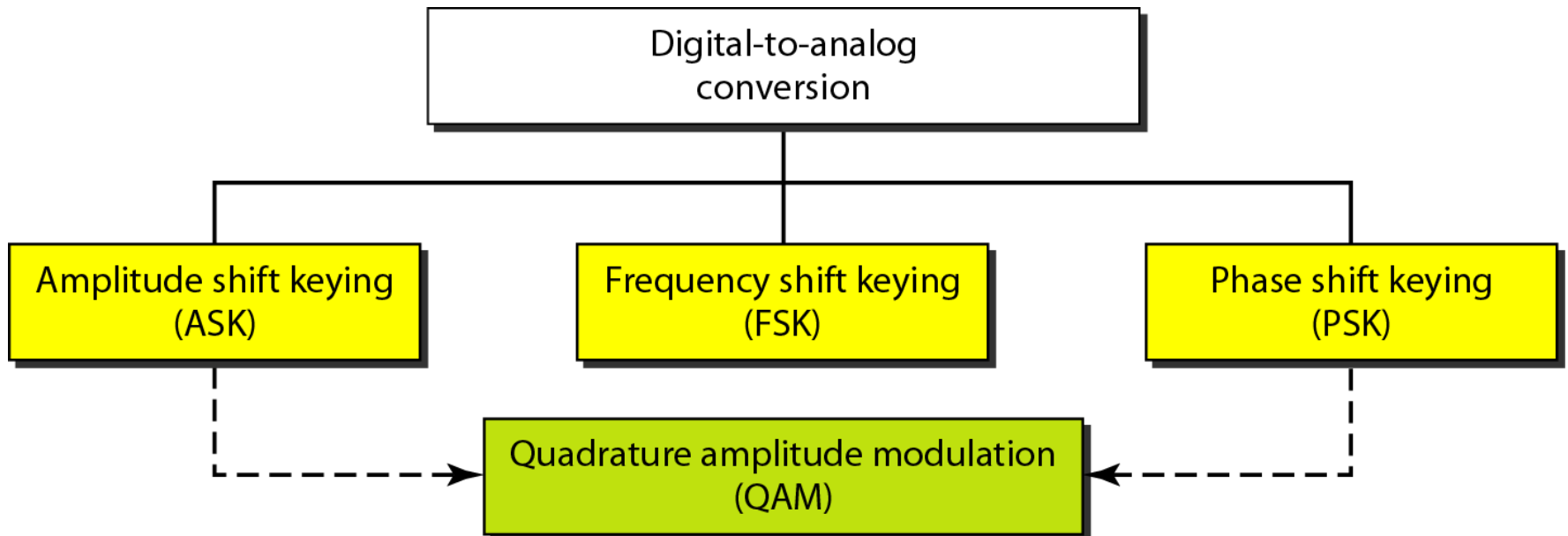


- Bandpass channels pass a range of frequencies around some center frequency f_c
 - Radio channels, telephone & DSL modems
- Digital modulators embed information into waveform with frequencies passed by bandpass channel
- Sinusoid of frequency f_c is centered in middle of bandpass channel
- Modulators embed information into a sinusoid

Digital To Analog Modulation



- process of changing one of the characteristics of an analog signal (typically a sinewave) based on the information in a digital signal
 - sinewave is defined by three characteristics (amplitude, frequency, and phase) ⇒ digital data (binary 0 and 1) can be represented by varying any of the three
 - **application**: transmission of digital data over telephone wire (modem)



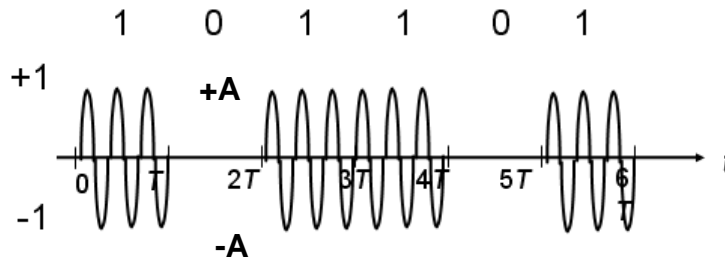
Amplitude Modulation



ASK — strength of the carrier signal is varied to represent binary 1 or 0

- both frequency and phase remain constant while the amplitude changes
- commonly, one of the amplitudes is zero

$$s(t) = \begin{cases} A_0 \cos(2\pi f_c t), & \text{binary 0} \\ A_1 \cos(2\pi f_c t), & \text{binary 1} \end{cases} = \begin{cases} 0, & \text{binary 0} \\ A \cos(2\pi f_c t), & \text{binary 1} \end{cases}$$

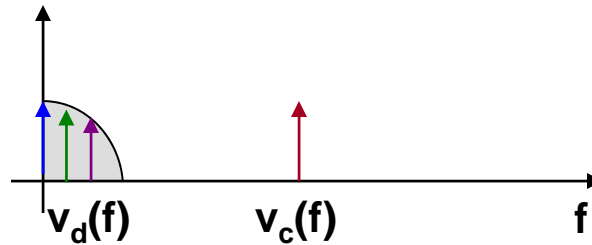
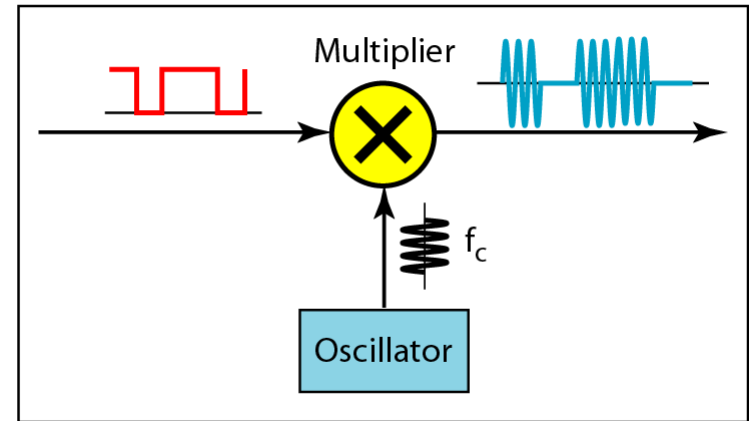
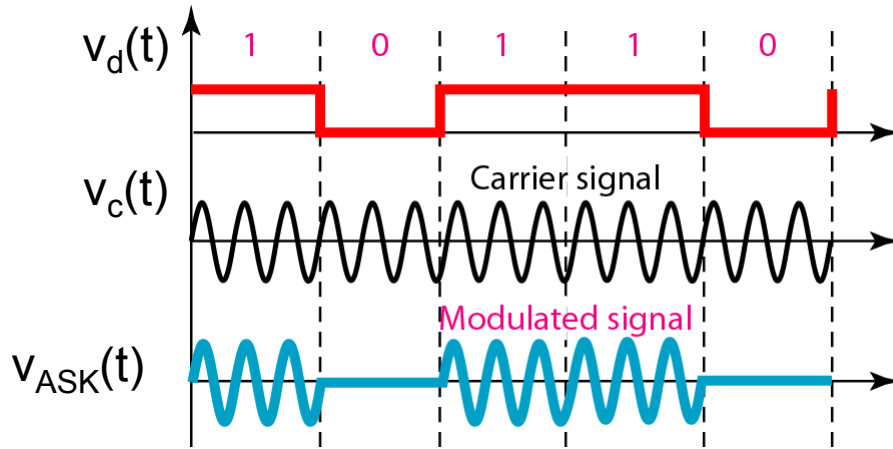


- **demodulation:** only the presence or absence of a sinusoid in a given time interval needs to be determined
- **advantage:** simplicity
- **disadvantage:** ASK is very susceptible to noise interference – noise usually (only) affects the amplitude, therefore ASK is the modulation technique most affected by noise
- **application:** ASK is used to transmit digital data over optical fiber

Amplitude Modulation (Cont.)



Example [ASK]



How does the frequency spectrum of $v_{ASK}(t)$ look like!?

Amplitude Modulation (Cont.)



ASK-Modulated Signal: Frequency Spectrum

$$\cos A \cdot \cos B = \frac{1}{2}(\cos(A - B) + \cos(A + B))$$

Carrier signal: $v_c(t) = \cos(2\pi f_c t) = \cos(\omega_c t)$, where $2\pi f_c = \omega_c$

Digital signal: $v_d(t) = A \cdot \left[\frac{1}{2} + \frac{2}{\pi} \cos \omega_0 t - \frac{2}{3\pi} \cos 3\omega_0 t + \frac{2}{5\pi} \cos 5\omega_0 t - \dots \right]$
(unipolar!!!)

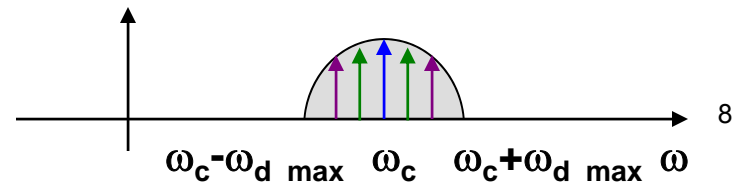
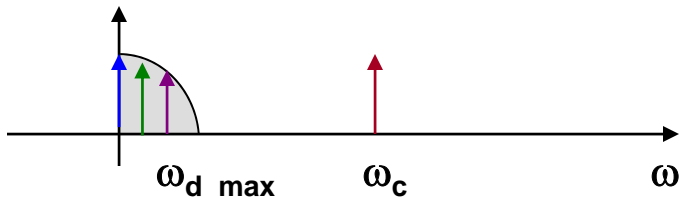
Modulated signal: $v_{ASK}(t) = v_c(t) \cdot v_d(t) =$

$$= \cos \omega_c t \cdot \left[\frac{1}{2} + \frac{2}{\pi} \cos \omega_0 t - \frac{2}{3\pi} \cos 3\omega_0 t + \frac{2}{5\pi} \cos 5\omega_0 t - \dots \right] =$$

$$= \frac{1}{2} \cos \omega_c t + \frac{2}{\pi} \cos \omega_c t \cdot \cos \omega_0 t - \frac{2}{3\pi} \cos \omega_c t \cdot \cos 3\omega_0 t + \dots =$$

$$= \frac{1}{2} \cos \omega_c t + \frac{1}{\pi} [\cos(\omega_c - \omega_0)t + \cos(\omega_c + \omega_0)t] -$$

$$- \frac{1}{3\pi} [\cos(\omega_c - 3\omega_0)t + \cos(\omega_c + 3\omega_0)t] + \dots$$

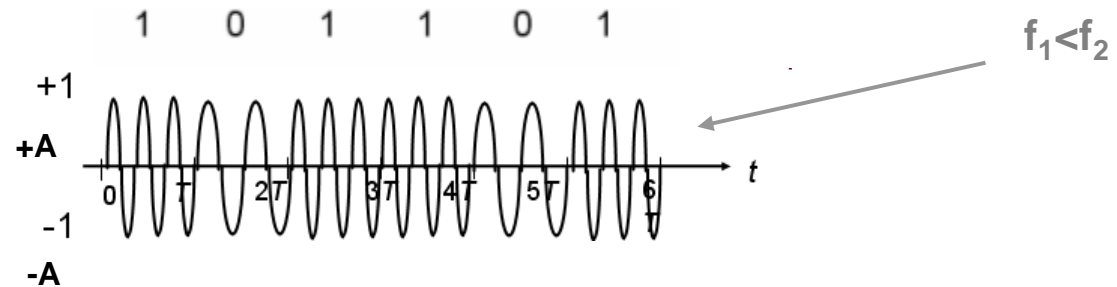


Frequency Modulation



- FSK** – frequency of the carrier signal is varied to represent binary 1 or 0
- both peak amplitude and phase remain constant during each bit interval

$$s(t) = \begin{cases} A\cos(2\pi f_1 t), & \text{binary 0} \\ A\cos(2\pi f_2 t), & \text{binary 1} \end{cases}$$

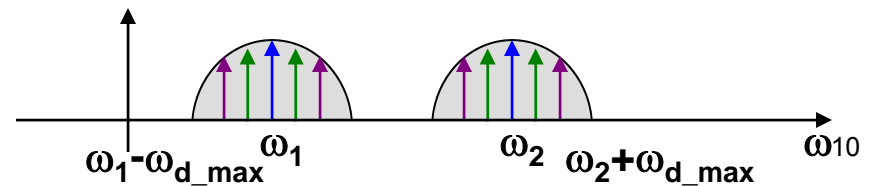
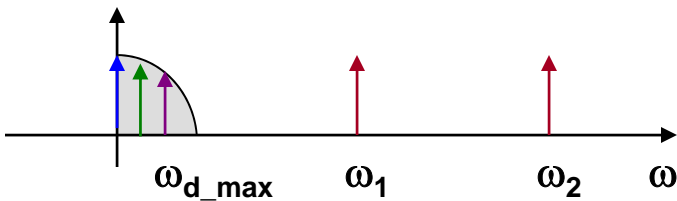
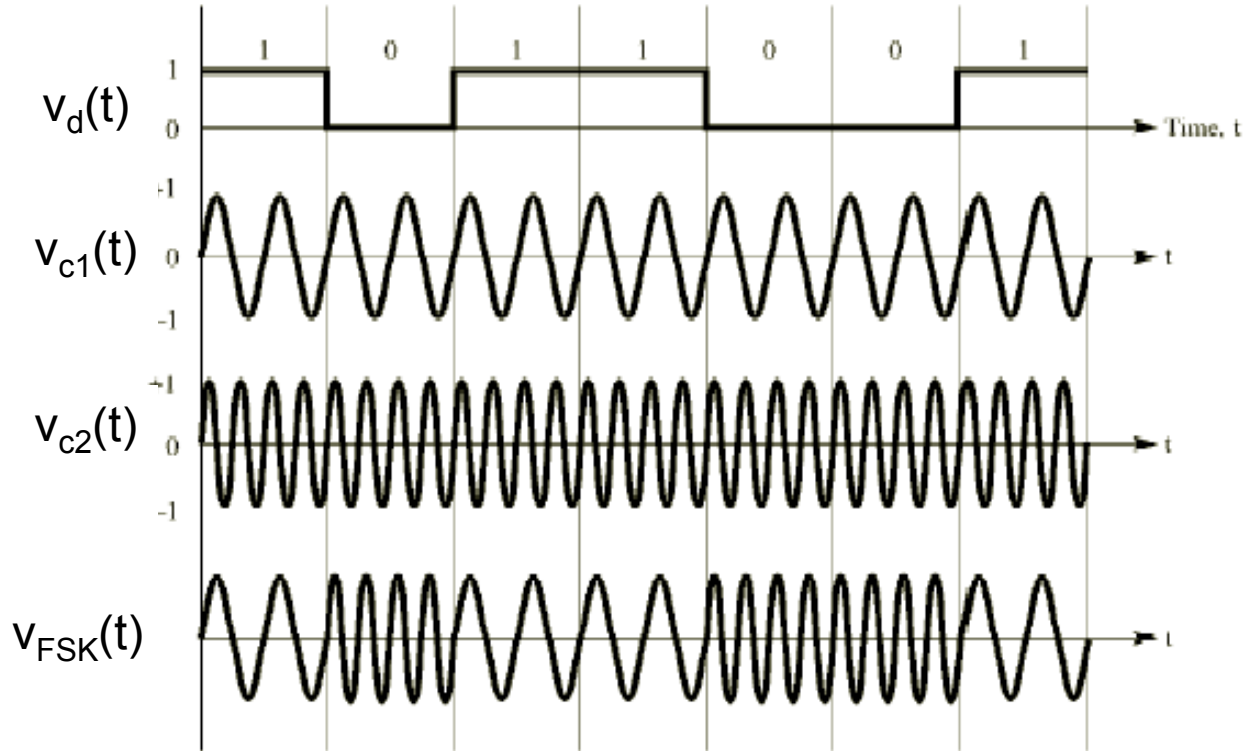


- **demodulation**: demodulator must be able to determine which of two possible frequencies is present at a given time
- **advantage**: FSK is less susceptible to errors than ASK – receiver is looking for specific frequency changes over a number of intervals, so voltage (noise) spikes can be ignored
- **disadvantage**: FSK spectrum is 2 x ASK spectrum
- **application**: over voice lines, in high-frequency radio transmission, etc. 9

Frequency Modulation (Cont.)



Example [FSK]



Frequency Modulation (Cont.)



FSK-Modulated Signal: Frequency Spectrum

Digital signal: $v_d(t)$ - modulated with ω_1 , and

$v_d'(t) = 1 - v_d(t)$ - modulated with ω_2

Modulated signal: $v_{FSK}(t) = \cos\omega_1 t \cdot v_d(t) + \cos\omega_2 t \cdot (1 - v_d(t)) =$

$$\begin{aligned} &= \cos\omega_1 t \cdot \left[\frac{1}{2} + \frac{2}{\pi} \cos\omega_0 t - \frac{2}{3\pi} \cos 3\omega_0 t + \frac{2}{5\pi} \cos 5\omega_0 t - \dots \right] + \\ &+ \cos\omega_2 t \cdot \left[\frac{1}{2} - \frac{2}{\pi} \cos\omega_0 t + \frac{2}{3\pi} \cos 3\omega_0 t - \frac{2}{5\pi} \cos 5\omega_0 t - \dots \right] = \\ &= \dots \\ &= \frac{1}{2} \cos\omega_1 t + \frac{1}{\pi} [\cos(\omega_1 - \omega_0)t + \cos(\omega_1 + \omega_0)t] - \\ &\quad - \frac{1}{3\pi} [\cos(\omega_1 - 3\omega_0)t + \cos(\omega_1 + 3\omega_0)t] + \dots + \\ &\frac{1}{2} \cos\omega_2 t - \frac{1}{\pi} [\cos(\omega_2 - \omega_0)t + \cos(\omega_2 + \omega_0)t] - \\ &\quad + \frac{1}{3\pi} [\cos(\omega_2 - 3\omega_0)t + \cos(\omega_2 + 3\omega_0)t] + \dots + \end{aligned}$$

Phase Modulation



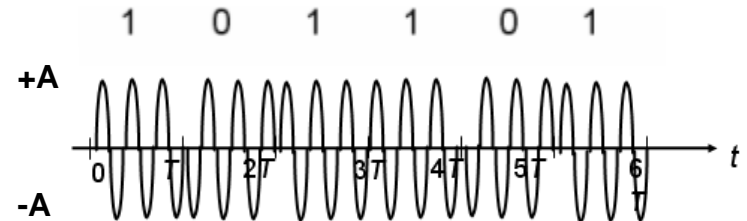
PSK – phase of the carrier signal is varied to represent binary 1 or 0

- peak amplitude and frequency remain constant during each bit interval
- example: binary 1 is represented with a phase of 0° , while binary 0 is represented with a phase of $180^\circ = \pi \text{rad}$ \Rightarrow **PSK is equivalent to multiplying the carrier signal by +1 when the information is 1, and by -1 when the information is 0**

2-PSK, or **Binary PSK**, since only 2 different phases are used.

$$s(t) = \begin{cases} A \cos(2\pi f_c t), & \text{binary 1} \\ A \cos(2\pi f_c t + \pi), & \text{binary 0} \end{cases}$$

$$s(t) = \begin{cases} A \cos(2\pi f_c t), & \text{binary 1} \\ -A \cos(2\pi f_c t), & \text{binary 0} \end{cases}$$

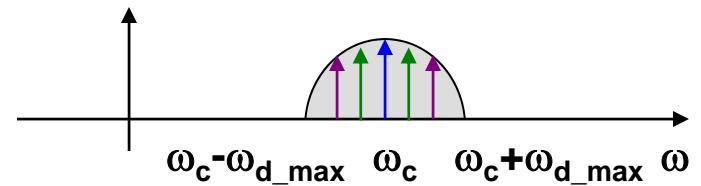
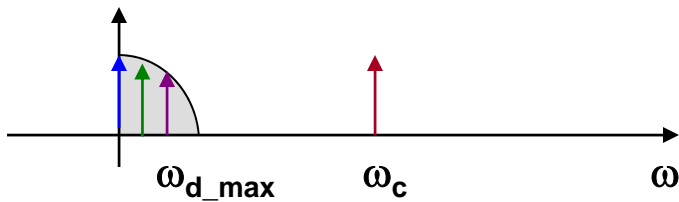
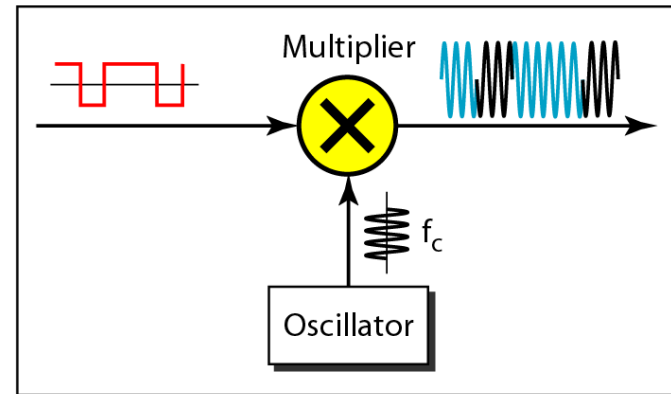
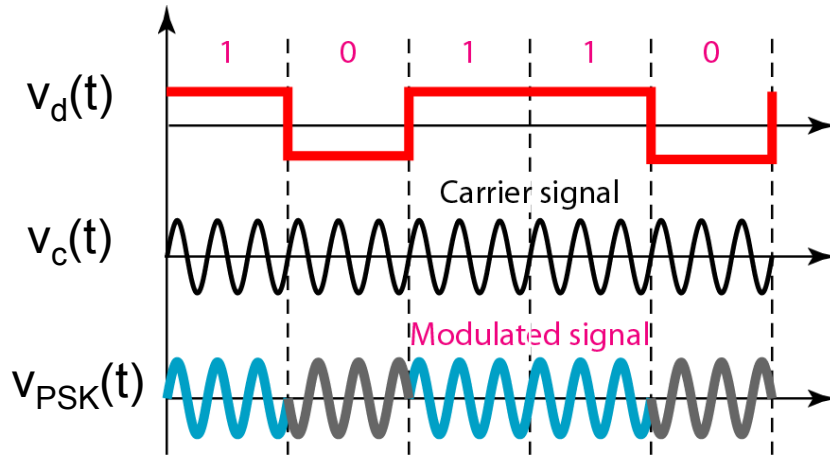


- **demodulation**: demodulator must be able to determine the phase of received sinusoid with respect to some reference phase
- **advantage**:
 - PSK is less susceptible to errors than ASK, while it requires/occupies the same bandwidth as ASK
 - more efficient use of bandwidth (higher data-rate) are possible, compared to FSK !!!
- **disadvantage**: more complex signal detection / recovery process, than in ASK and FSK

Phase Modulation (Cont.)



Example [PSK]



Phase Modulation (Cont.)

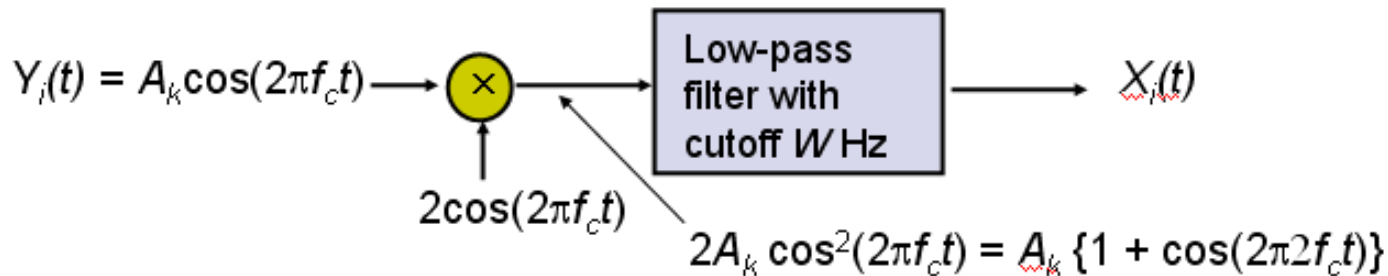


PSK Detection / Recovery – multiply the received / modulated signal $\pm A\cos(2\pi f_c t)$ by $2\cos(2\pi f_c t)$

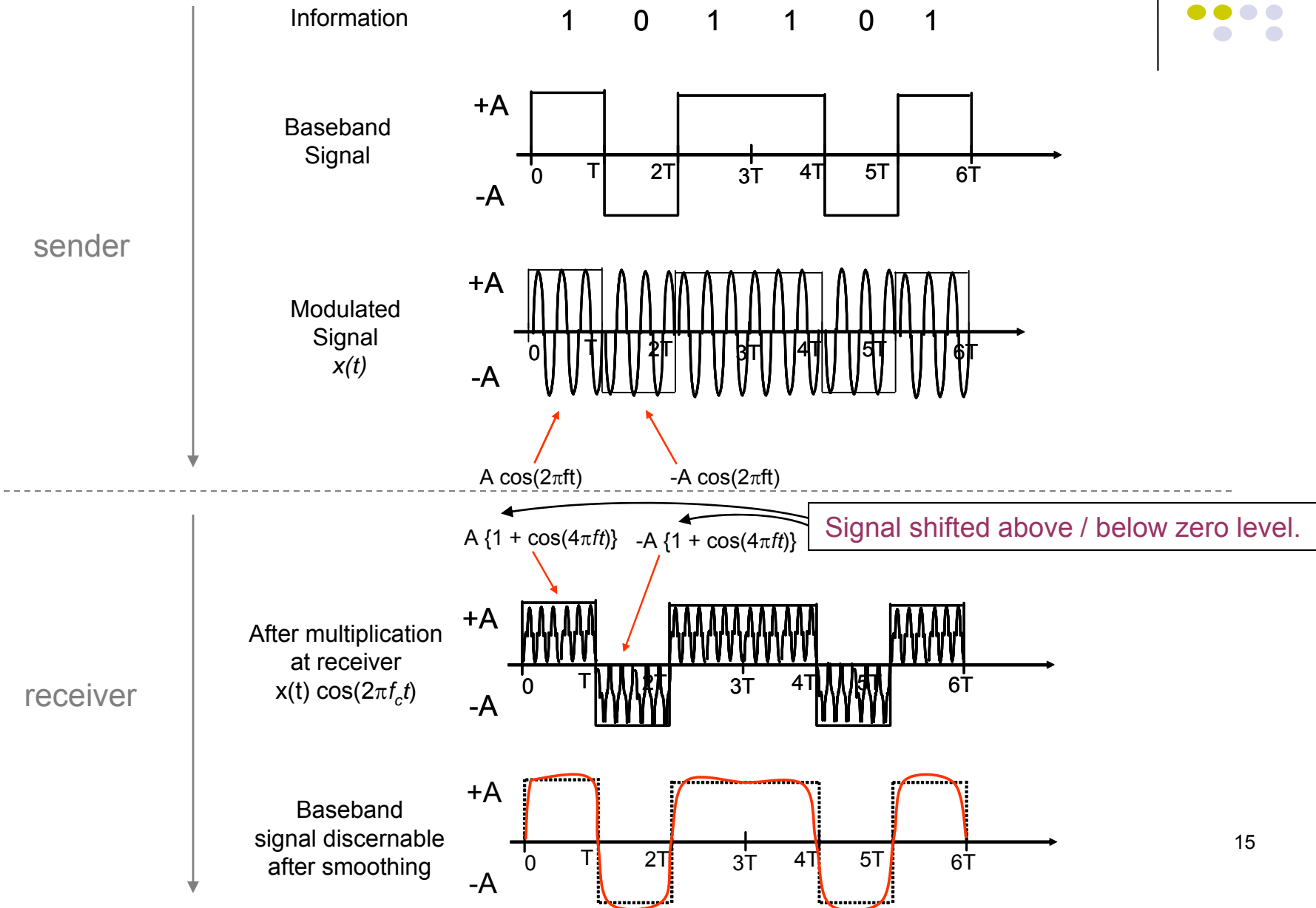
$$\cos^2 A = \frac{1}{2}(1 + \cos 2A)$$

- resulting signal
 $2A\cos^2(2\pi f_c t) = A[1 + \cos(4\pi f_c t)]$, binary 1
- $-2A\cos^2(2\pi f_c t) = -A[1 + \cos(4\pi f_c t)]$, binary 0

- by removing the oscillatory part with a low-pass filter, the original baseband signal (i.e. the original binary sequence) can be easily determined



Phase Modulation (Cont.)



Signaling rate and Transmission Bandwidth



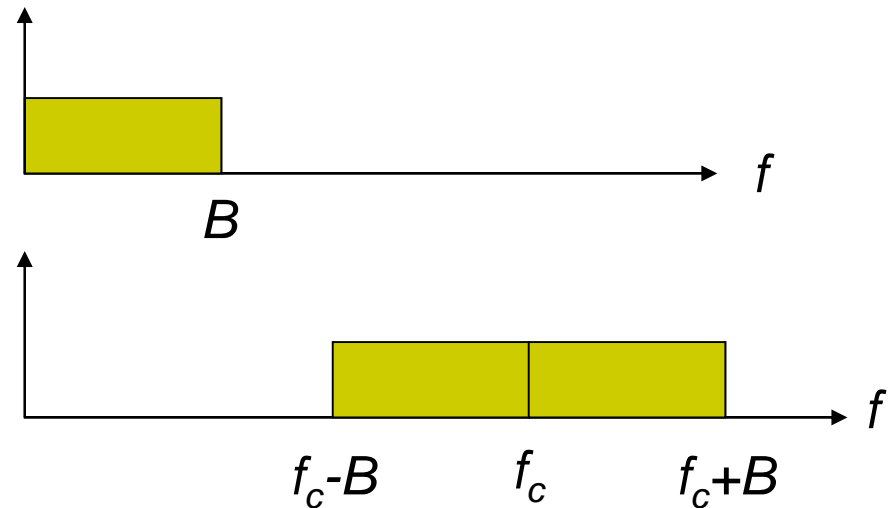
- Fact from modulation theory:

If

Baseband signal $x(t)$
with bandwidth B Hz

then

Modulated signal
 $x(t)\cos(2\pi f_c t)$ has
bandwidth $2B$ Hz



- If bandpass channel has bandwidth W_c Hz,
 - Then baseband channel has $W_c/2$ Hz available, so
 - modulation system supports $W_c/2 \times 2 = W_c$ pulses/second

Recall **baseband transmission system of bandwidth W_c [Hz] can theoretically support $2 W_c$ pulses/sec**

Phase Modulation (Cont.)



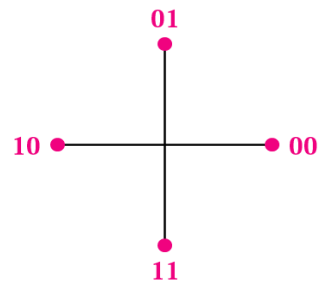
QPSK = 4-PSK

– PSK that uses phase shifts of $90^\circ = \pi/2$ rad \Rightarrow 4 different signals are generated, each representing 2 bits

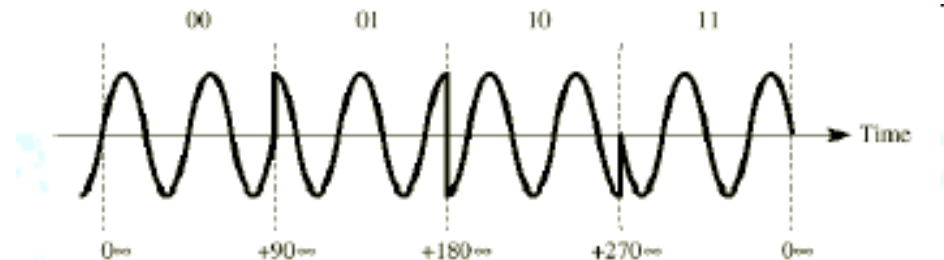
$$s(t) = \begin{cases} A\cos(2\pi f_c t), & \text{binary 00} \\ A\cos(2\pi f_c t + \frac{\pi}{2}), & \text{binary 01} \\ A\cos(2\pi f_c t + \pi), & \text{binary 10} \\ A\cos(2\pi f_c t + \frac{3\pi}{2}), & \text{binary 11} \end{cases}$$

Dibit	Phase
00	0
01	90
10	180
11	270

Dibit
(2 bits)



Constellation diagram



- **advantage:** higher data rate than in PSK (2 bits per bit interval), while bandwidth occupancy remains the same
- 4-PSK can easily be extended to 8-PSK, i.e. n-PSK
- however, higher rate PSK schemes are limited by the ability of equipment to distinguish small differences in phase

Quadrature Amplitude Modulation (QAM)

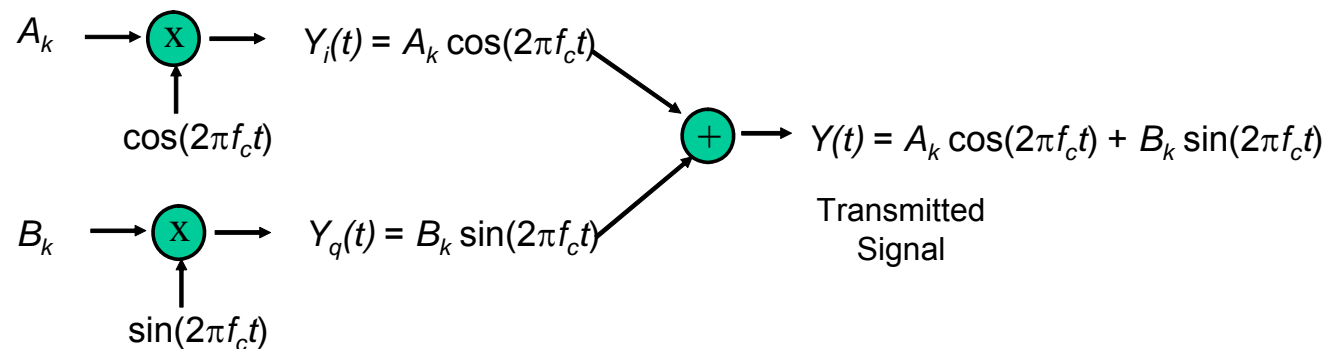
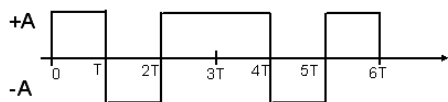


- uses “two-dimensional” signalling
 - original information stream is split into two sequences that consist of odd and even symbols, e.g. B_k and A_k

1 0 1 1 0 1 ...
 1 -1 1 1 -1 1 ...

B_1 A_1 B_2 A_2 B_3 A_3 ...

- A_k sequence (**in-phase component**) is modulated by $\cos(2\pi f_c t)$, while B_k sequence (**quadrature-phase component**) is modulated by $\sin(2\pi f_c t)$
- composite signal $A_k \cos(2\pi f_c t) + B_k \sin(2\pi f_c t)$ is sent through the channel

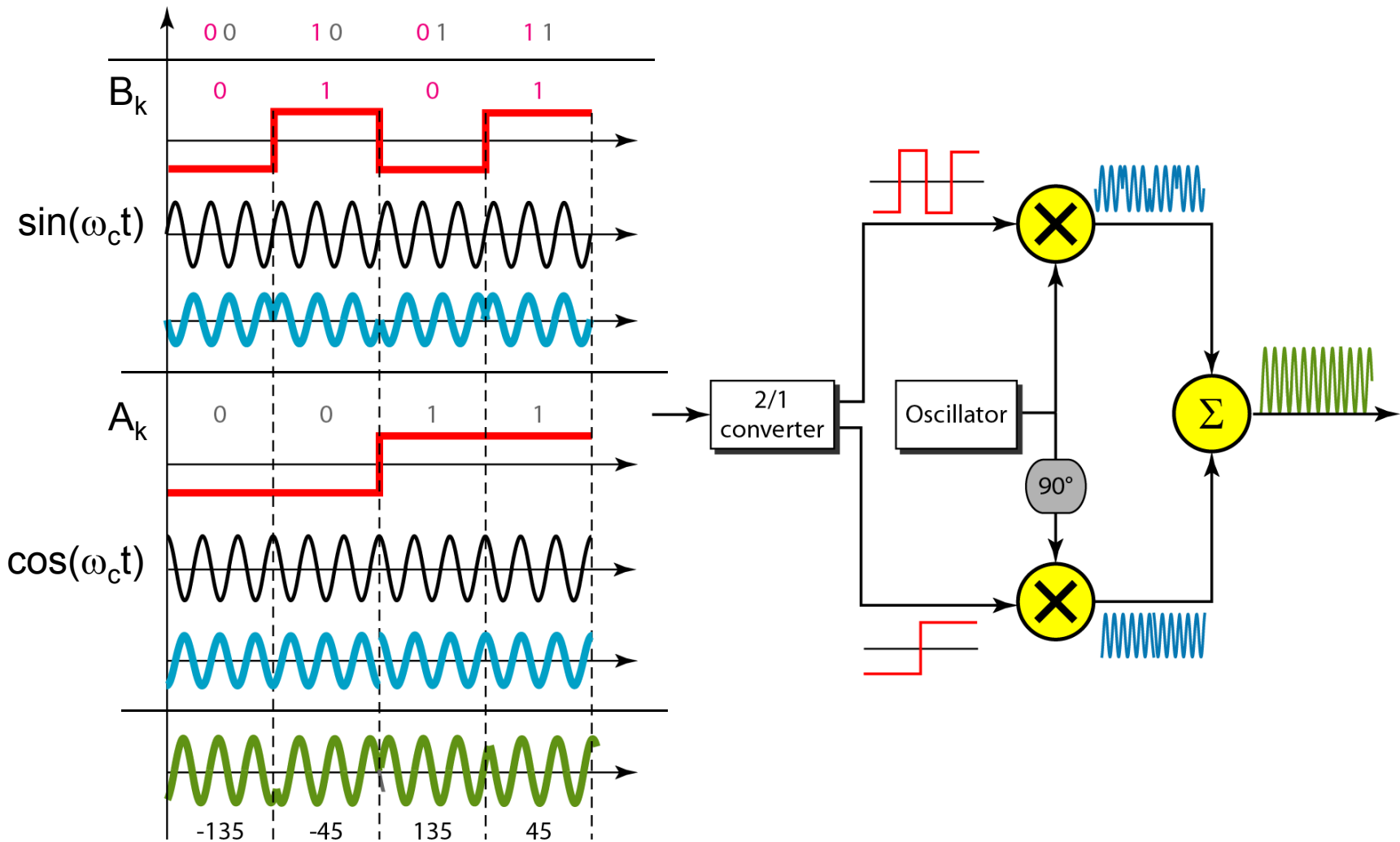


- **advantage:** data rate = 2 bits per bit-interval!

QAM (Cont.)



Example [QAM]

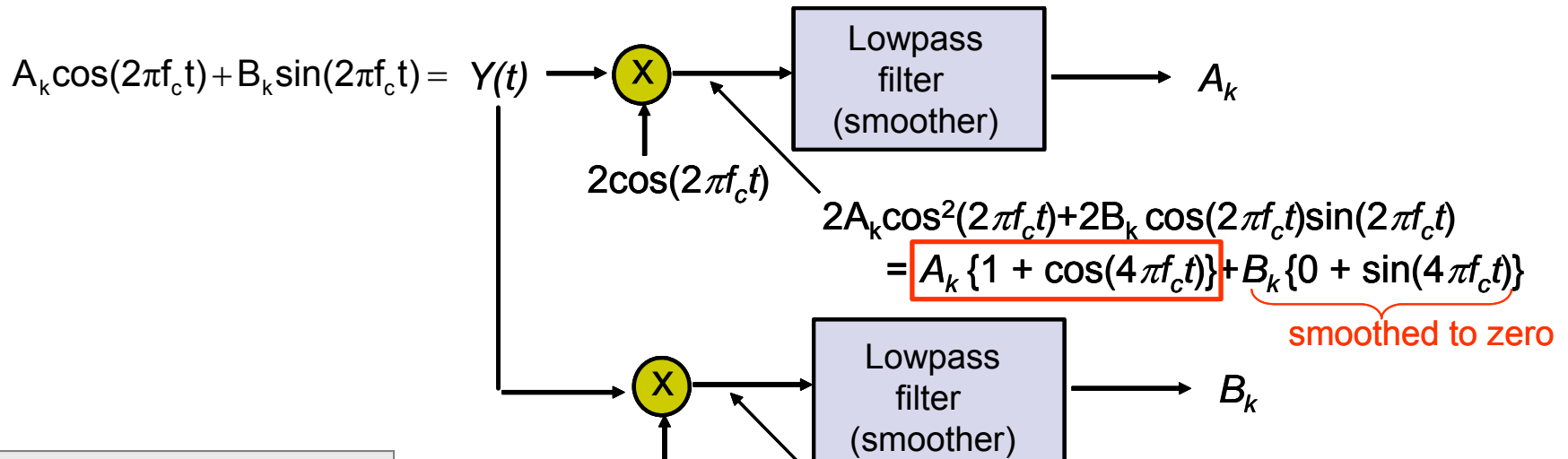


QAM (Cont.)



QAM Demodulation

- by multiplying $Y(t)$ by $2 \cdot \cos(2\pi f_c t)$ and then low-pass filtering the resultant signal, sequence A_k is obtained
- by multiplying $Y(t)$ by $2 \cdot \sin(2\pi f_c t)$ and then low-pass filtering the resultant signal, sequence B_k is obtained



$$\begin{aligned}
 &2A_k \cos^2(2\pi f_c t) + 2B_k \cos(2\pi f_c t) \sin(2\pi f_c t) \\
 &= A_k \{1 + \cos(4\pi f_c t)\} + B_k \{0 + \sin(4\pi f_c t)\}
 \end{aligned}$$

smoothed to zero

$$\begin{aligned}
 &2B_k \sin^2(2\pi f_c t) + 2A_k \cos(2\pi f_c t) \sin(2\pi f_c t) \\
 &= B_k \{1 - \cos(4\pi f_c t)\} + A_k \{0 + \sin(4\pi f_c t)\}
 \end{aligned}$$

smoothed to zero

$$\cos^2(A) = \frac{1}{2}(1 + \cos(2A))$$

$$\sin^2(A) = \frac{1}{2}(1 - \cos(2A))$$

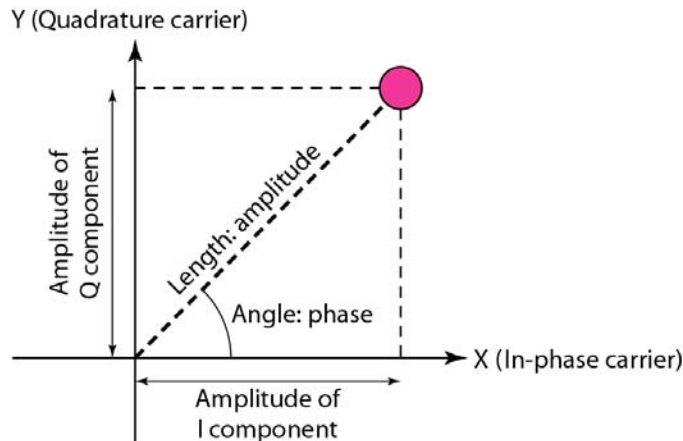
$$\sin(2A) = 2\sin(A)\cos(A)$$

Signal Constellations

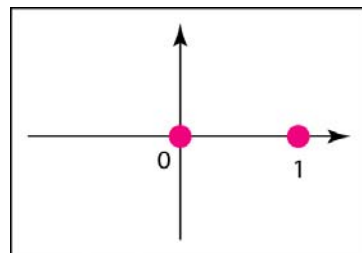


Constellation Diagram

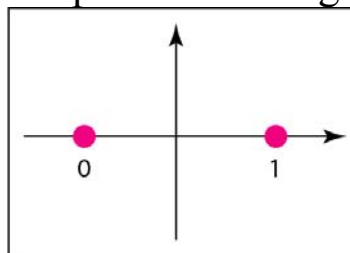
- used to represent possible symbols that may be selected by a given modulation scheme as points in 2-D plane



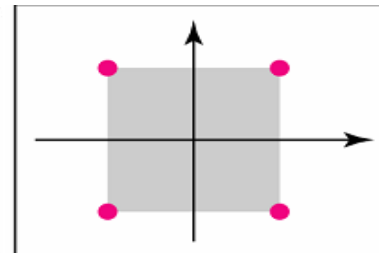
- X-axis is related to in-phase carrier: $\cos(\omega_c t)$
 - the projection of the point on the X-axis defines the peak amplitude of the in-phase component
- Y-axis is related to the quadrature carrier: $\sin(\omega_c t)$
 - the projection of the point on the Y-axis defines the peak amplitude of the quadrature component
- the length of the line that connects the point to the origin is the peak amplitude of the signal element (combination of X and Y components)
- the angle the line makes with the X-axis is the phase of the signal element



a. ASK (OOK)



b. BPSK



b. 4-QAM

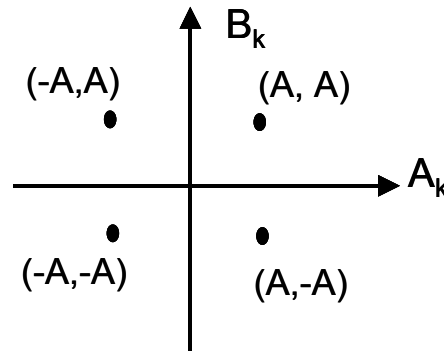
QAM (Cont.)



QAM cont. – QAM can also be seen as a combination of ASK and PSK

$$Y(t) = A_k \cos(2\pi f_c t) + B_k \sin(2\pi f_c t) = \left(A_k^2 + B_k^2\right)^{\frac{1}{2}} \cos\left(2\pi f_c t + \tan^{-1} \frac{B_k}{A_k}\right)$$

4-level QAM



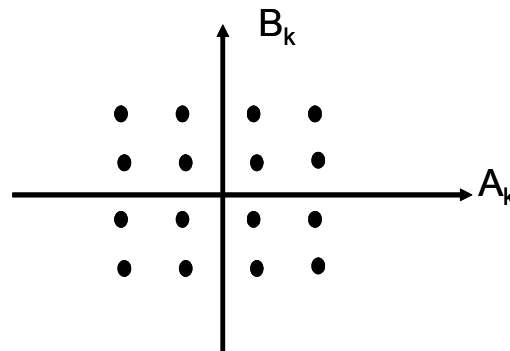
QAM (Cont.)



16-level QAM – the number of bits transmitted per T [sec] interval can be further increased by increasing the number of levels used

- in case of 16-level QAM, A_k and B_k individually can assume 4 different levels: -1, -1/3, 1/3, 1
- data rate: **4 bits/pulse \Rightarrow 4W bits/second**

$$Y(t) = A_k \cos(2\pi f_c t) + B_k \sin(2\pi f_c t) = \left(A_k^2 + B_k^2\right)^{\frac{1}{2}} \cos\left(2\pi f_c t + \tan^{-1} \frac{B_k}{A_k}\right)$$



A_k and B_k individually can take on 4 different values; the resultant signal can take on (only) 3 different values!!!

In QAM various combinations of amplitude and phase are employed to achieve higher digital data rates.

Amplitude changes are susceptible to noise \Rightarrow the number of phase shifts used by a QAM system is always greater than the number of amplitude shifts.