

## L12: Basic Baseband Digital Comms



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## Outline

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- Basic digital communicator blocks
- Nyquist pulses
- Channel capacity

## Underlying Idea

- Attempting to send a sequence of digits through a continuous channel



- Not easy...
- ...modularize the design

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## The Information Theory Perspective

- Exchanging sequences of **discrete symbols**
- source coding
  - increase redundancy
- channel coding
  - error correction (add redundancy)
- line coding
  - map bits to symbols



data                      symbols

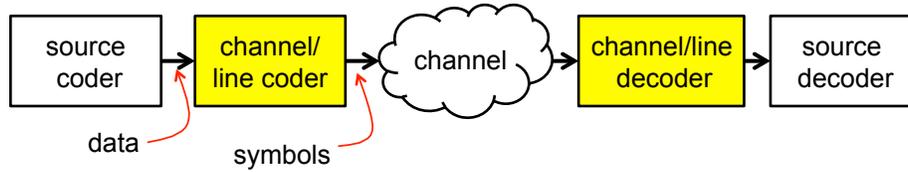
0	—	-1	00	—	-3	00	—	-1
1	—	+1	01	—	-1	01	—	-j
			10	—	+1	10	—	+1
			11	—	+3	11	—	+j

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## Line Coding Examples



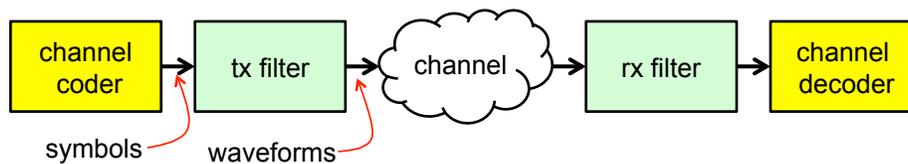
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## Continuous-Time Communication

- Transmit filter (TX filter)
- Map symbols into analog waveforms that match channel response
  - bandwidth
  - location (baseband/passband)



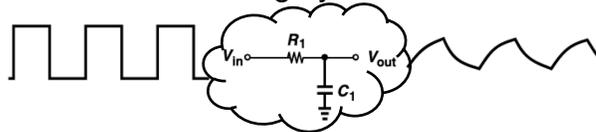
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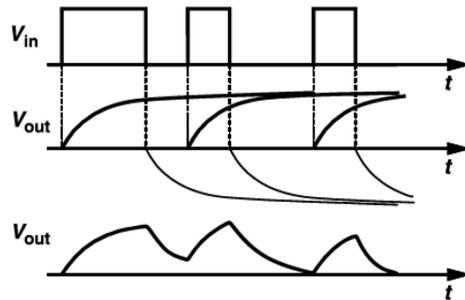
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## Baseband Channel Problems

- Appreciate need to design your waveforms to channel



- Prolonged rise and fall times corrupt energies of adjacent symbols: **intersymbol interference (ISI)**



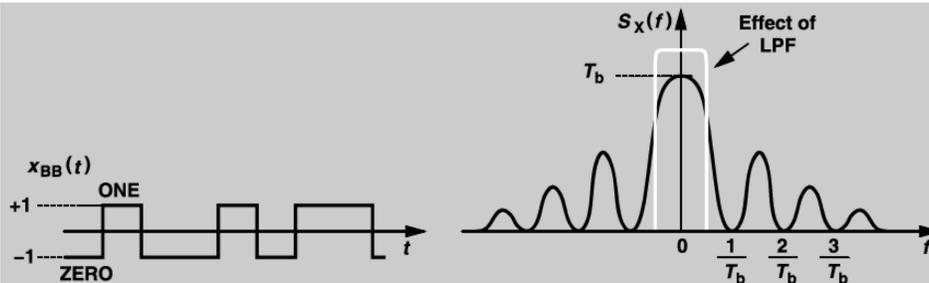
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## A Spectral Perspective

- Random bit patterns  $\text{sinc}^2$  spectrum



- Need **extremely wide LPF** to pass this without distortion
- Otherwise deal with **ISI** (intersymbol interference)

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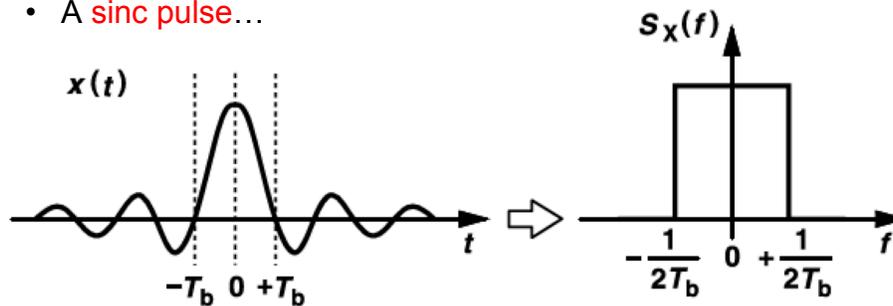
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## Pulse Shaping, Sinc Pulses

- What is the ideal bandlimited pulse shape for the tx filter to produce?
- A **sinc pulse**...



- ...has spectrum that is neatly confined (to  $\pm 1/2T_b$ )
- And there's another important advantage of this...

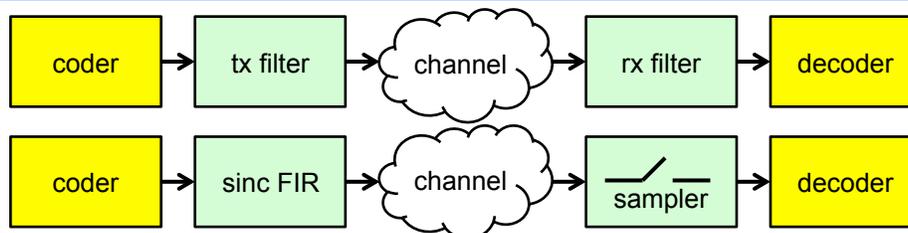
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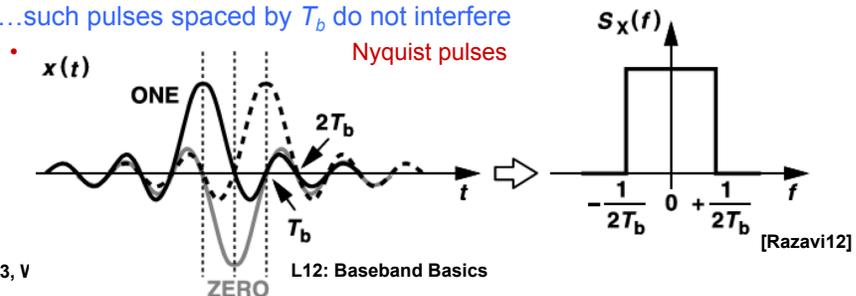
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## Simple Baseband Modulator (Pulse Shaper)



- Pulse goes through zero at equally spaced intervals of  $T_b$
- ...such pulses spaced by  $T_b$  do not interfere



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## Example of Composite Waveform

Three Nyquist pulses shown separately

- $+s(t)$
- $+s(t-T)$
- $-s(t-2T)$

Composite waveform

$$r(t) = s(t) + s(t-T) - s(t-2T)$$

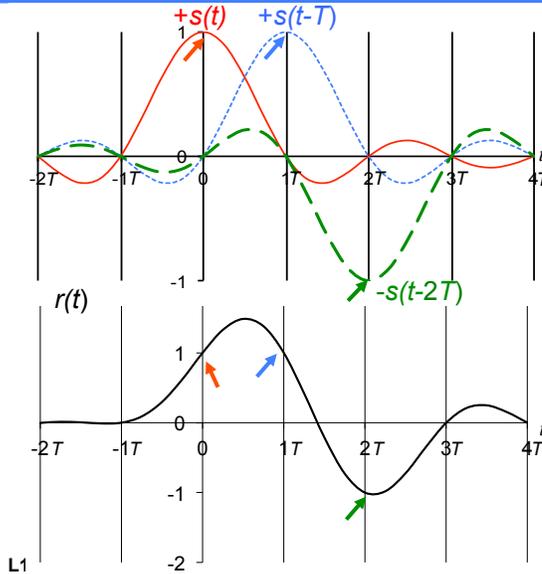
Samples at  $kT$

$$r(0) = s(0) + s(-T) - s(-2T) = +1$$

$$r(T) = s(T) + s(0) - s(-T) = +1$$

$$r(2T) = s(2T) + s(T) - s(0) = -1$$

Zero ISI at sampling times  $kT$

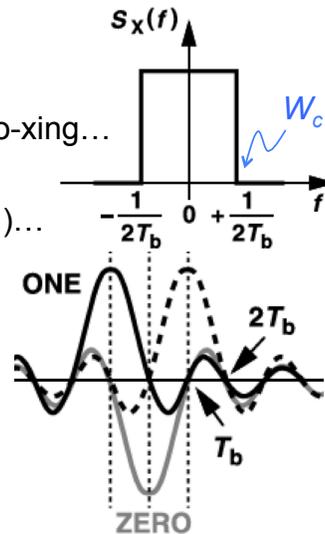


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## Achievable Symbol Rate

- For brickwall channel...
  - $W_c$ : stop frequency
- For sinc pulses with  $2T_b$  between zero-xing...
  - spectrum limited to  $1/2T_b$
- Therefore to prevent distortion (no ISI)...
  - make:  $1/2T_b = W_c$
- Therefore for  $W_c$  channel...
  - can send: symbols per second =  $2 \cdot W_c$
- E.g.: 1 MHz channel allows 2 MSps



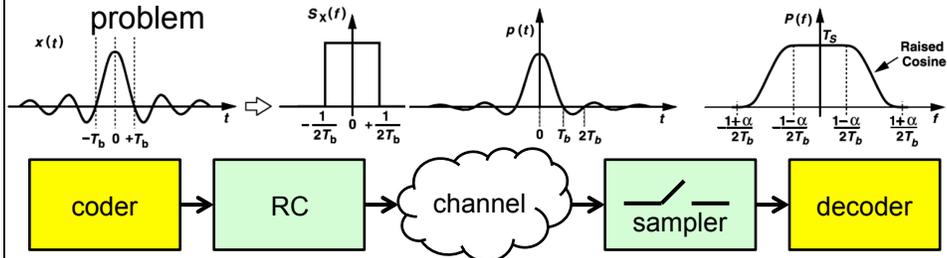
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## Raised Cosine (RC)

- Simple sinc has a lot of temporal energy away from main pulse
  - Sampling error can lead to lots of corruption
- **Raised cosine** (RC) waveform trades bandwidth for this problem



- excess-bandwidth/roll-off factor,  $\alpha$ 
  - 10-20%

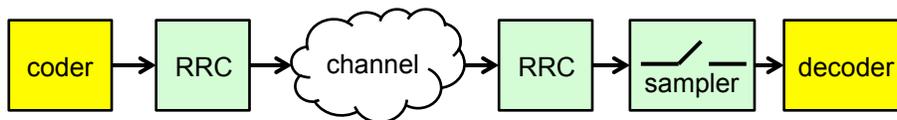
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## Root Raised Cosine (RRC)

- Very important for receiver to filter-out noise
  - minimize corruption to sampler
- Split up RC for **simultaneous** noise filtering and pulse shaping



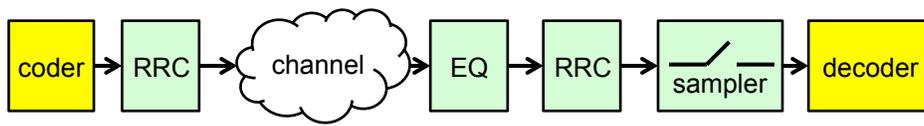
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## Equalization & Matched Filtering

- What if channel not a perfect brickwall or RC filter?
  - our previous assumptions would not match such a general channel and hence be distorted
- Compensate for channel with another filter: **equalizer** (EQ)



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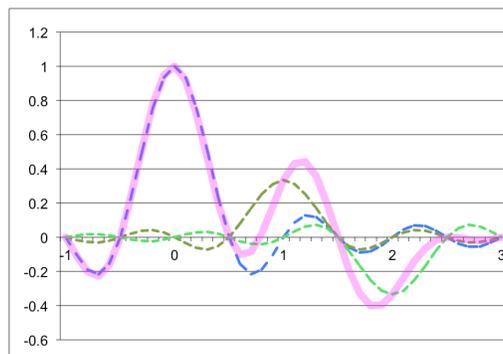
## Achievable Bit Rate, $R$

- Account for **alphabet size:  $M$**

- binary,  $M = 2$
- multilevel  $M > 2$
- 4-level:
 

00	—	-1
01	—	-1/3
10	—	+1/3
11	—	+1

- $R = 2W_c \cdot \log_2(M)$
- For example:
  - $W_c = 1$  MHz
  - $M = 4$
  - $R = 4$  Mbps (2 MSps)



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## Spectral Efficiency

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- What is the bit-rate relative to the spectrum used:

$$v = \frac{\text{bit-rate}}{\text{bandwidth}}$$

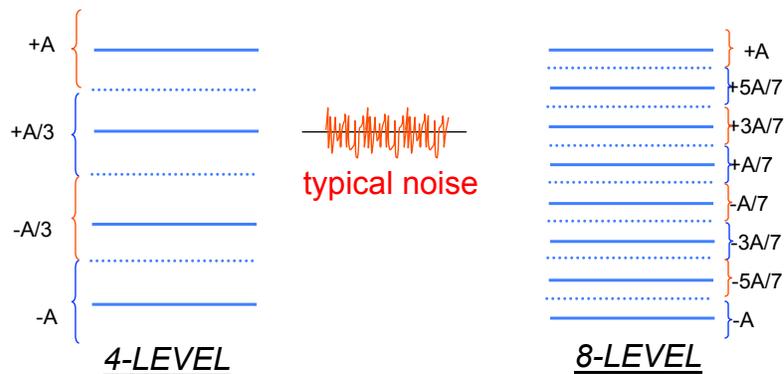
$$v = \frac{2W_c \cdot \log_2(M)}{W_c} \left[ \frac{\text{bits}}{\text{s} \cdot \text{Hz}} \right]$$

- Simple 2-level baseband modulation 2 bits/s•Hz
- or **2 symbols/s•Hz**

## Noise Limits Accuracy

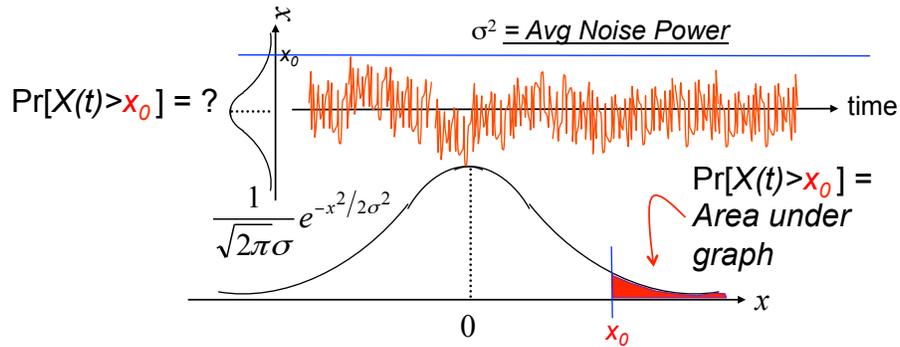
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- Receiver makes decision based on transmitted pulse level + noise
- **Error rate** depends on relative value of noise and level spacing
- Large (positive or negative) noise values can cause wrong decision
- Noise level impacts 8-level signaling more than 4-level signaling



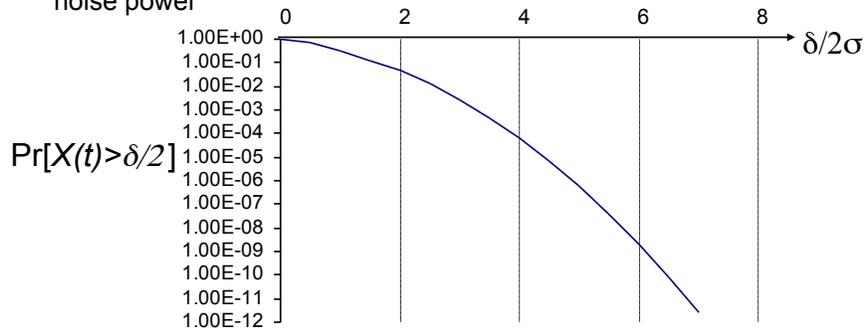
## Noise Distribution

- Noise is characterized by **probability density** of amplitude samples
- **Likelihood** that certain amplitude occurs
- **Thermal electronic** noise is inevitable (due to vibrations of electrons)
- Noise distribution is **Gaussian** (bell-shaped) as below



## Probability of Error

- Error occurs if noise value **exceeds certain magnitude**
- Prob. of large values **drops quickly** with Gaussian noise
- **Target probability of error** achieved by designing system so separation between signal levels is appropriate relative to average noise power



## Shannon Channel Capacity

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### Shannon Channel Capacity:

- The maximum reliable transmission rate over an ideal channel with bandwidth  $W_c$  Hz, with Gaussian distributed noise, and with SNR  $S/N$  is

$$C = W_c \log_2(1 + S/N) \text{ bits per second}$$

- Reliable means error rate can be made arbitrarily small by proper coding

## Example

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- Consider a 3 kHz channel with 8-level signaling. Compare bit rate to channel capacity at 20 dB SNR
- 3 kHz telephone channel with 8 level signaling  
Bit rate =  $2 \cdot 3000$  pulses/sec  $\cdot 3$  bits/pulse = 18 kbps
- 20 dB SNR means  $10 \log_{10}(S/N) = 20$   
Implies  $S/N = 100$
- Shannon Channel Capacity is then  
 $C = 3000 \log(1 + 100) = 19,963$  bits/second