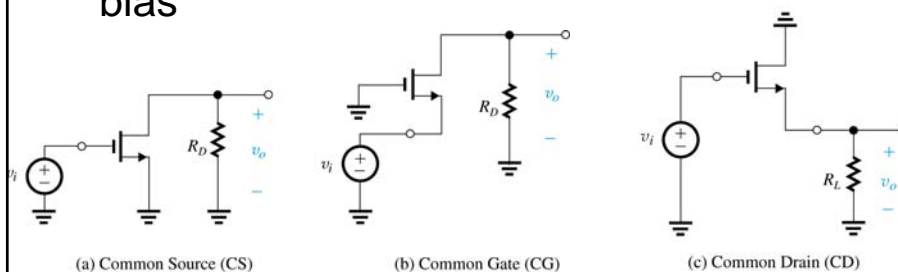


MOSFET Amplifier Configuration

- Single stage
- The signal is fed to the amplifier represented as v_{sig} with an internal resistance R_{sig} .
- MOSFET is represented by its small signal model.
- Generally interested in gain, input and output resistance (overall amplifier circuit not only the small signal model).

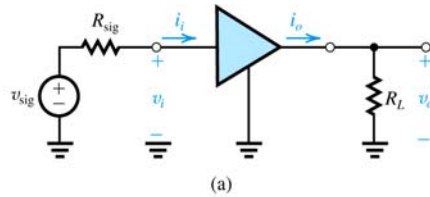
MOSFET Amplifier Configuration

- Considering only the small signal not the bias



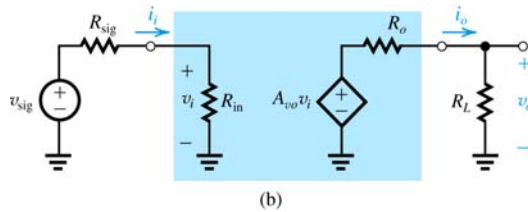
Characterizing Amplifiers

- Find gain, input and output resistance



$$R_{in} = \frac{v_i}{i_i}$$

$$A_{VO} = \left. \frac{v_o}{v_i} \right|_{R_L = \infty}, A_V = \frac{v_o}{v_i}$$



$$G_v = \frac{v_o}{v_{sig}} \text{ Overall voltage gain}$$

Amplifier Configuration

- Common Source
- Common Source with a source resistance
- Common gate
- Common drain or voltage follower

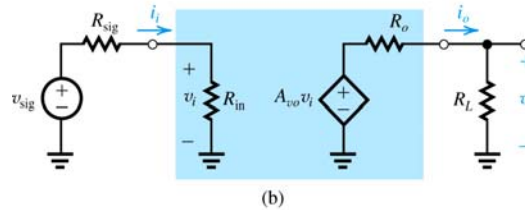
Amplifiers

$$v_o = A_{vo} v_i \frac{R_L}{R_L + R_o}$$

$$v_i = v_{sig} \frac{R_{in}}{R_{in} + R_{sig}}$$

$$A_v = A_{vo} \frac{R_L}{R_L + R_o}$$

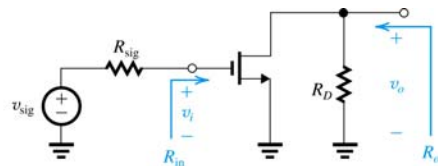
$$G_v = \frac{v_o}{v_{sig}} = \frac{R_{in}}{R_{in} + R_{sig}} A_{vo} \frac{R_L}{R_L + R_o}$$



Common Source

- Most widely used configuration
- In multistage amplifiers, the bulk of the gain is from common source.
- The source is grounded, making it common between input and output.
- We can use hybrid π model.

Common Source



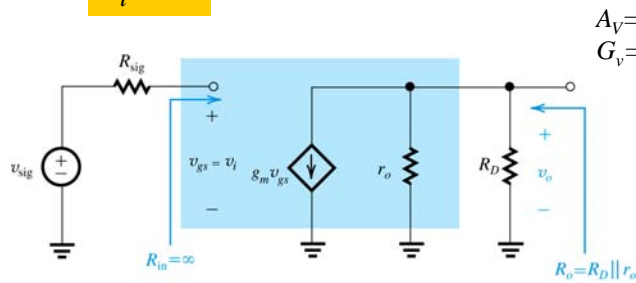
$R_i = \infty$

(a)

For R_o , set $v_i=0$

$R_o = r_o \parallel R_D$

$A_{vo} = -g_m (r_o \parallel R_D)$

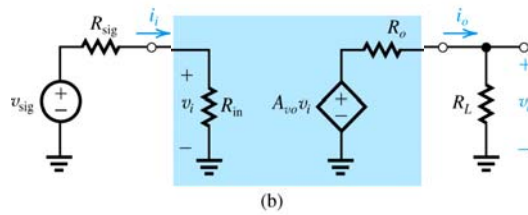


$A_v = ?$
 $G_v = ?$

(b)

Example

CS at $I_D=0.25 \text{ mA}$, $V_{OV}=0.25\text{V}$, $R_D=20\text{k} \Omega$,
Signal source has a resistance of $100\text{k}\Omega$, and $20\text{k} \Omega$ load
Find:
 R_{in} , A_{vo} , R_o , A_v , and G_v

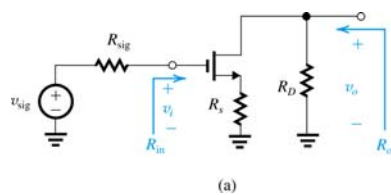


(b)

Common Source with Source R

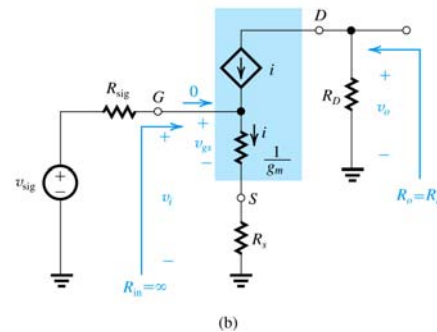
- For simplicity, r_o is not included.
- No effect on discrete implementation, not so for IC's
- R_s provides a negative feedback to control the magnitude of the signal to prevent nonlinear distortion.
- Also reduces the voltage gain and extends the useful bandwidth.

Common Source with Source R



$$v_{gs} = v_i \frac{1/g_m}{1/g_m + R_S}$$

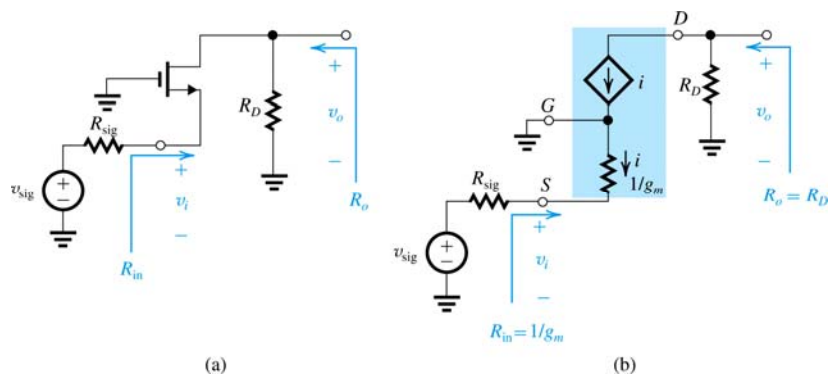
$$v_{gs} = \frac{v_i}{1 + g_m R_S}$$



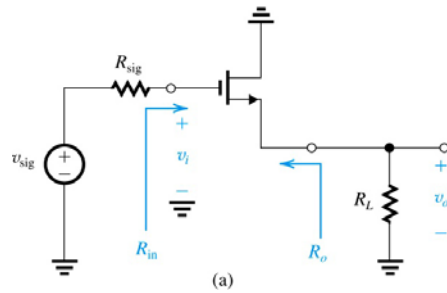
Example

In the previous example, $v_{in}=50$ mV, the output was 1V
 Now, we have $v_{sig} = 0.2$ V, we want to modify the circuit to keep v_{gs} unchanged, what value we should use for R_s

Common Gate Amplifier

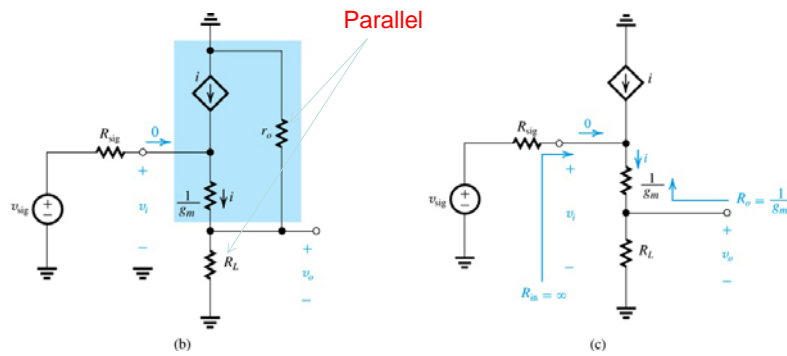


Common Drain Amplifier – Voltage Follower



Since there is a resistance R_L connected to the source, it is easier to use the T-model

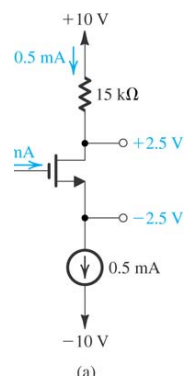
Common Drain– Voltage Follower



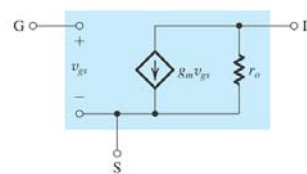
$R_i = \infty$

Comparison

	CS	CS+RS	CG	CD
Rin	∞	∞	$\frac{1}{g_m}$	∞
Rout	$R_D \parallel r_o$	R_D	R_D	$1/g_m$
G	$-g_m(R_D \parallel R_L \parallel r_o)$	$A_v = \frac{g_m(R_D \parallel R_L)}{1 + g_m R_S}$	$G_v = \frac{(R_D \parallel R_L)}{1/g_m + R_{sis}}$	$G_v = \frac{A_v R_L}{1/g_m + R_L}$



$V_{OV} = 1V$
 $V_{GS} = 2.5V$



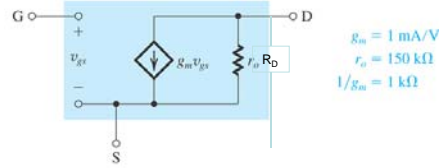
$g_m = 1 \text{ mA/V}$
 $r_o = 150 \text{ k}\Omega$
 $1/g_m = 1 \text{ k}\Omega$

$V_{DD} = V_{SS} = 10V, I = 0.5 \text{ mA}, R_G = 4.7M\Omega, R_D = 15K\Omega,$
 $V_t = 1.5V, k_n = 1 \text{ mA/V}^2, V_A = 75V$

Find $V_{OV}, V_{GS}, V_G, V_S, V_D, g_m, r_o$

What is the max. possible voltage swing at drain and the MOSFET remains in saturation?

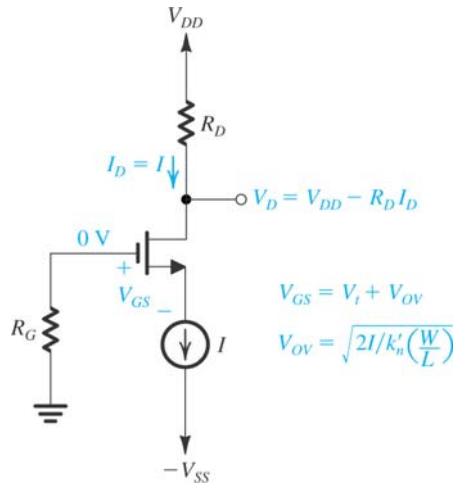
Find R_{in} , A_{vo} , R_o , G_v with and without r_o . $R_{sig}=100\text{K}\Omega$ and $R_L=15\text{K}\Omega$



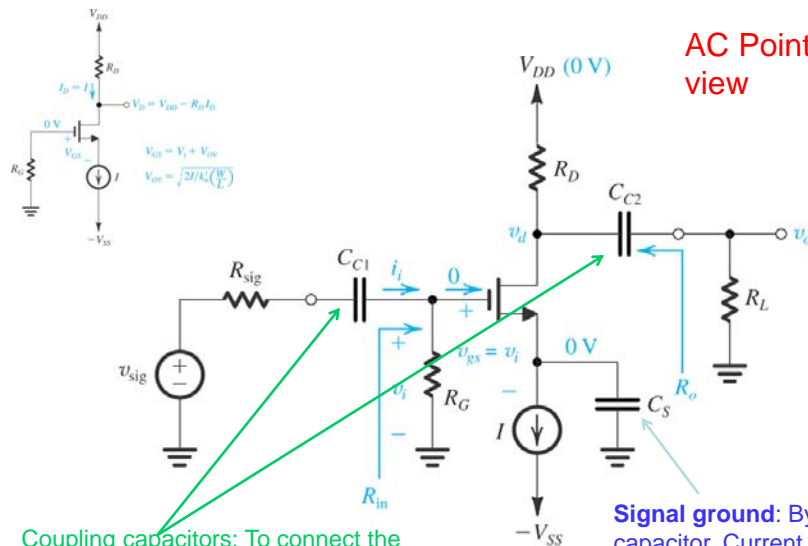
Discrete CMOS Amplifiers

- MOS are mostly used in IC's
- However, we present practical circuits for discrete implementation.
- Good if you are building a circuit on a breadboard (in the lab).

Common Source



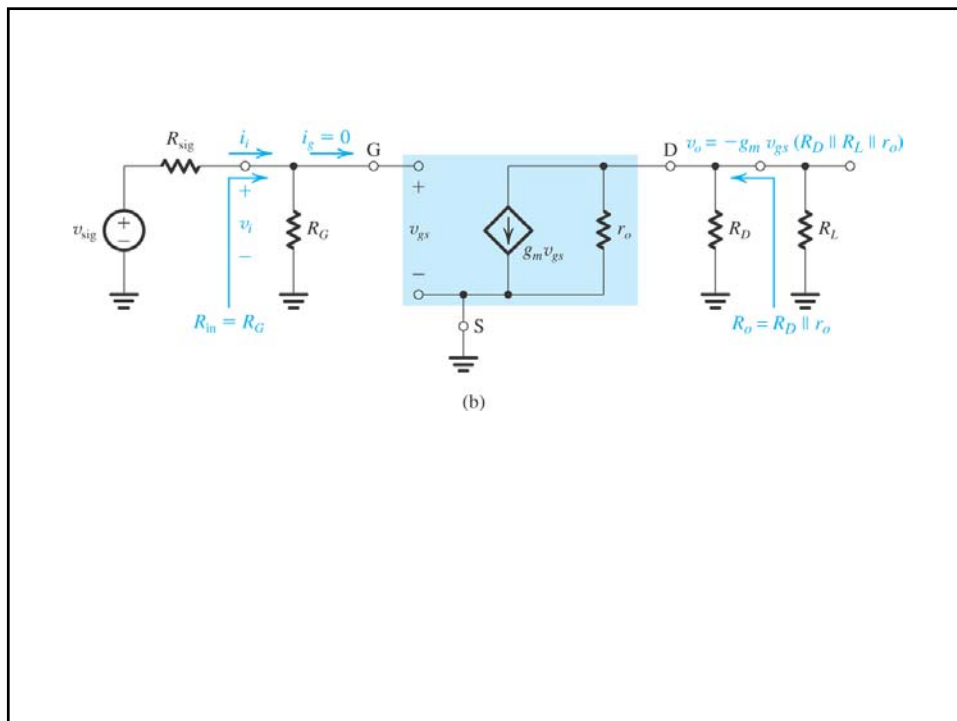
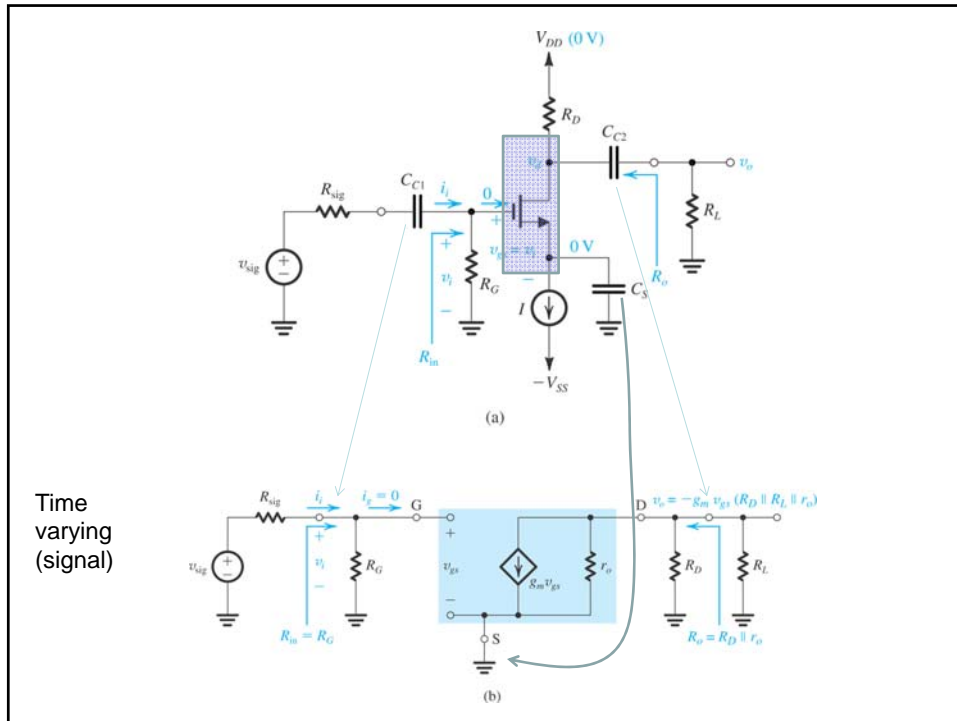
- DC Analysis
- Can use any biasing technique we studied
- Calculate the bias (Q) point



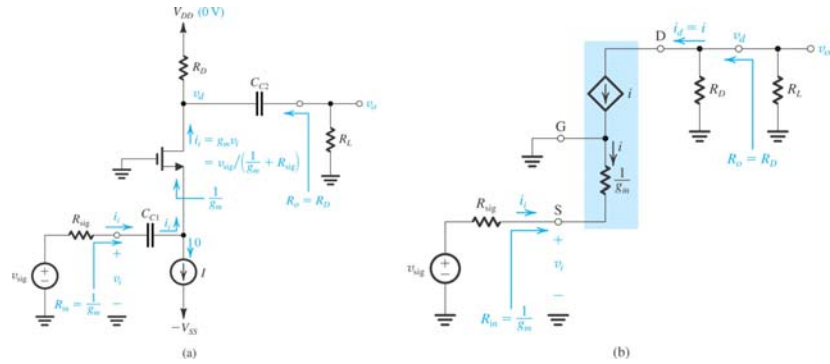
AC Point of view

Coupling capacitors: To connect the input and output without disturbing the bias.
 C is large enough to have very low (zero) impedance at signal frequency.

Signal ground: Bypass capacitor. Current source is an open circuit, provide a path to ground (source is ground for signal).



Common Gate



Voltage Follower

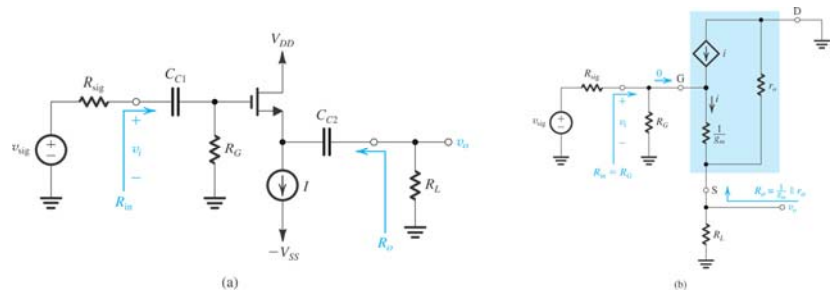


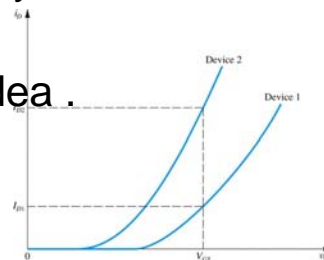
Figure 5.60 (a) A source-follower amplifier. (b) Small-signal, equivalent-circuit model.

Biassing in MOS Amplifiers

- How to choose the operating point?
- Want a stable Q-point (known I_D and V_{DS}) to ensure operation in the saturation region.

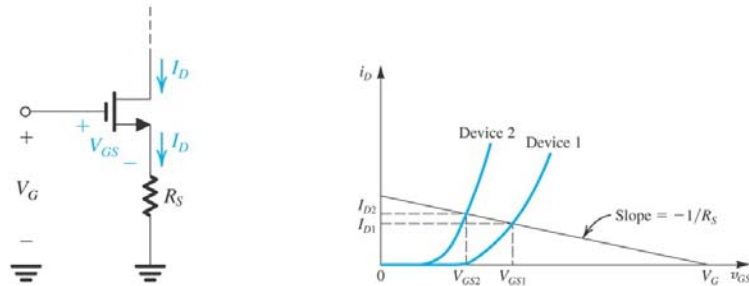
Biassing -- Fixing V_{GS}

- I_D depends on μ , C_{ox} , W/L and V_t , and V_{GS}
- C_{ox} , V_{GS} (even W/L) can vary across devices of the same type.
- Constant V_{GS} Not a good idea .
- μ, C_{ox} are a f(t)



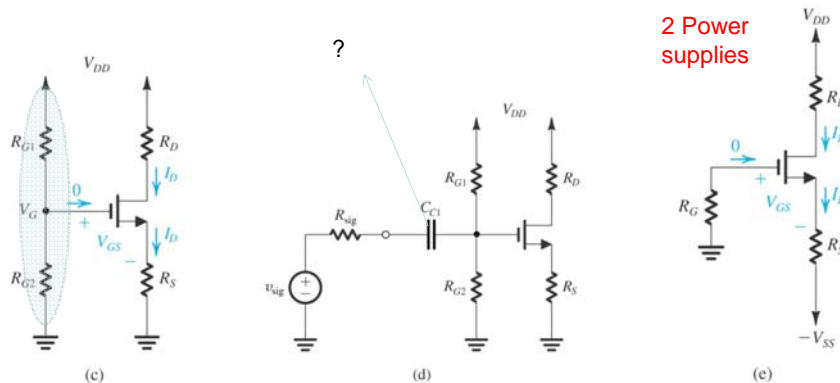
Biasing – Fixing V_G and R_S

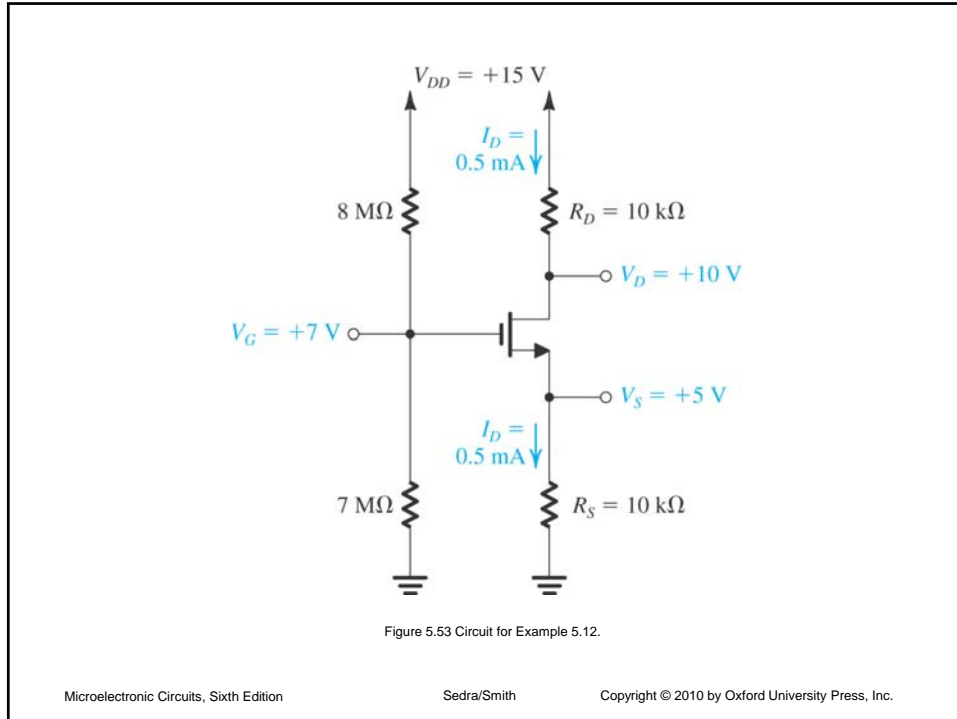
- R_S provides a negative feedback to stabilize I_D



Biasing – Fixing V_G and R_S

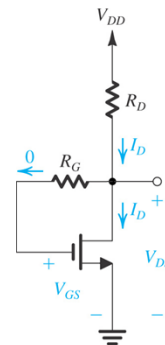
- Uses one power supply
- What is the effect on input resistance when you add v_{gs} signal





Biasing – D-to-G Resistor

- $V_{GS} = V_{DS} = V_{DD} - I_D R_D$
- $V_{DD} = V_{GS} + I_D R_D$
- Provides a feedback resistor to stabilize I_D



Biasing – Constant Current Source

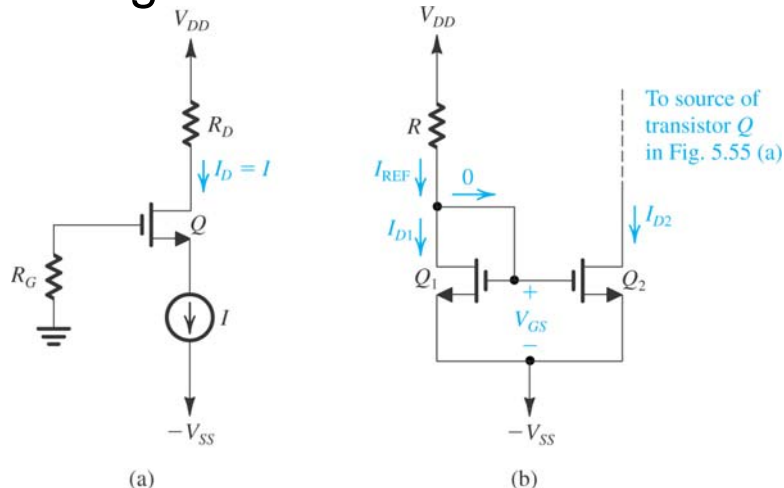


Figure 5.55 (a) Biasing the MOSFET using a constant-current source I . (b) Implementation of the constant-current source I using a current mirror.

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Frequency Response

