

## L12: Equalization



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# Outline

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- Intro to Equalization
- ZFE
  - Zero-Forcing Equalizer
- MMSE
  - Minimum-Mean-Squared Equalizer
- DFE
  - Decision-Feedback Equalizer

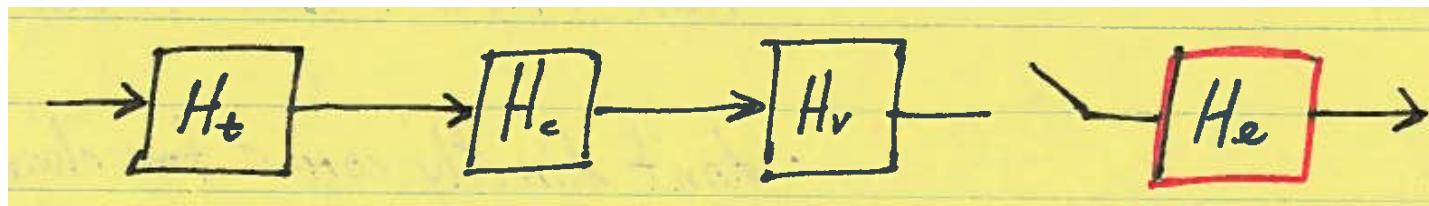
## 1.2 Overview

- To maximize SNR and eliminate ISI set

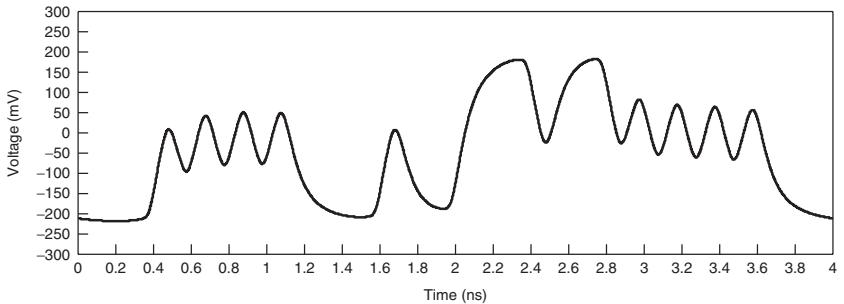
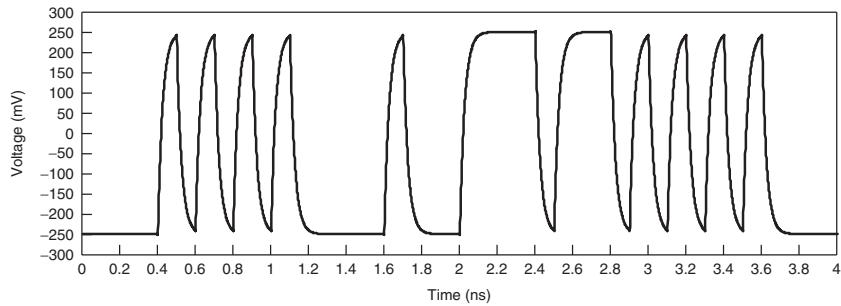
$$|H_r(f)| = \frac{\alpha |P_r|^{\frac{1}{2}}}{G_n^{\frac{1}{4}} |H_c|^{\frac{1}{2}}}$$

$$|H_t| = \frac{(A/\alpha) |P_r|^{\frac{1}{2}} G_n^{\frac{1}{4}}}{|H_c|^{\frac{1}{2}}}$$

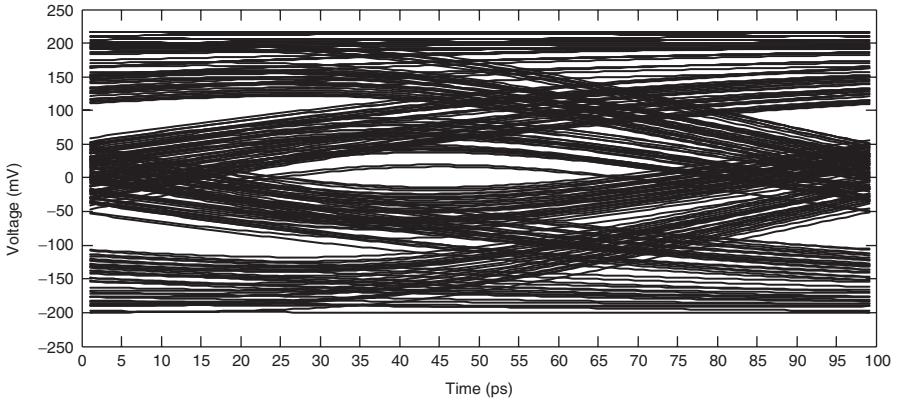
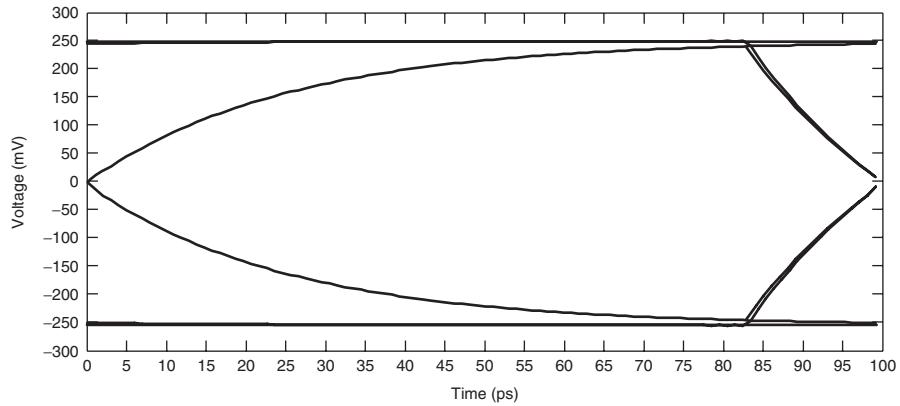
- These may be hard to implement
  - Let TX and RX focus on pulse shaping & noise
- Introduce another filter to focus on ISI
  - The equalizer



# Overview



- Eye diagrams



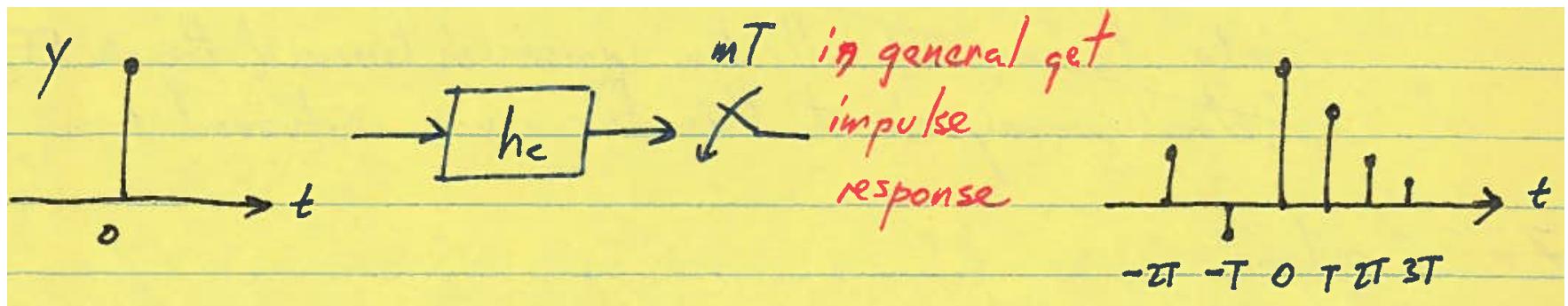
## 12.2 Equalizer Types

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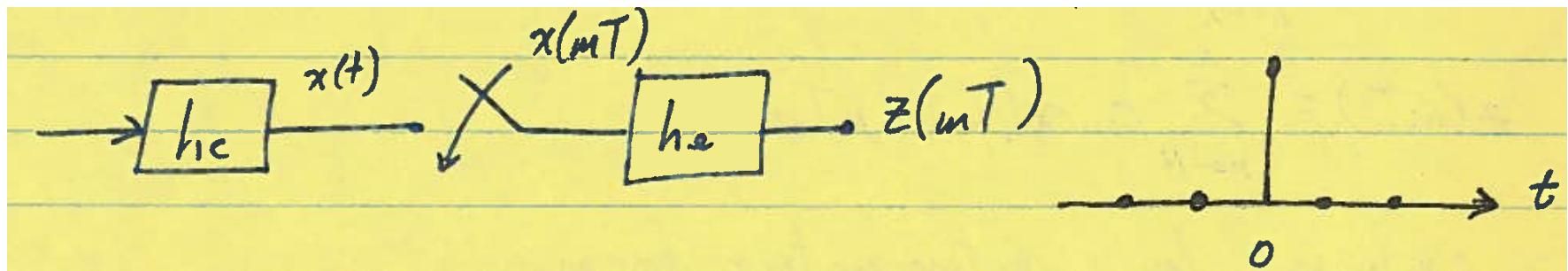
- Filter-Based
  - Linear (Transversal Filters)
    - ZFE
    - MMSE
  - Non-Linear
    - DFE
- Other
  - Maximum-Likelihood Sequence Estimation (MLSE)
    - Viterbi

## 12.3 Zero-Forcing Equalizer (ZFE)

- If this is the impulse response before your detector...

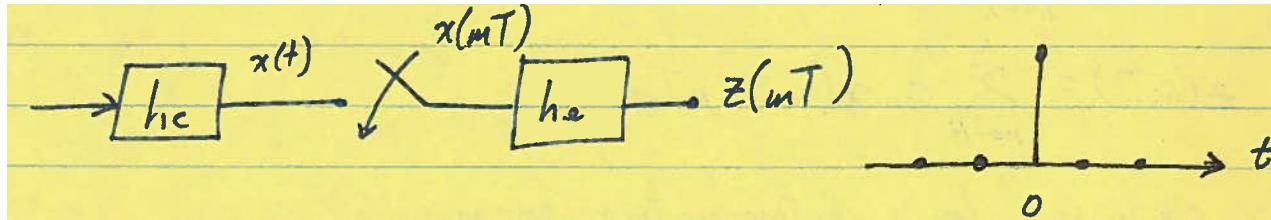


- The ZFE attempts to achieve...



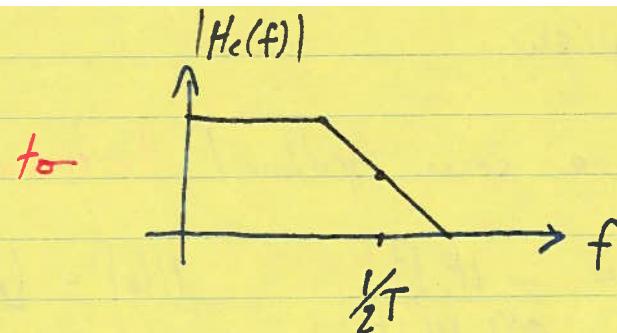
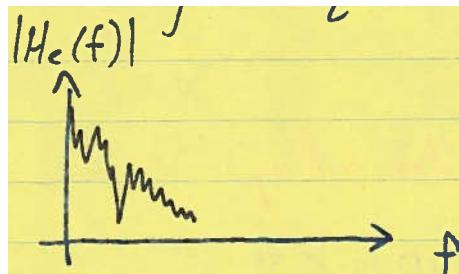
# ZFE Response

- This response is effectively achieved by...



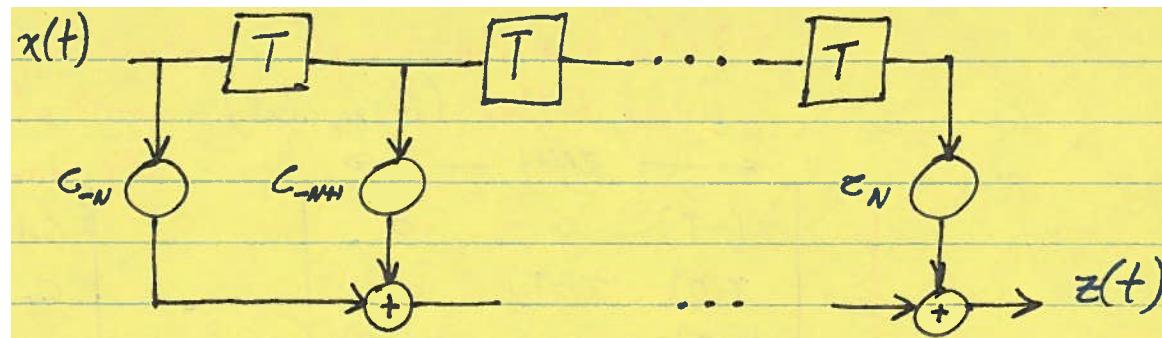
- ...inverting the channel

$$H_e(f) = \frac{1}{H_c(f)} = \frac{1}{|H_c(f)|} \cdot e^{-j\theta_c(f)}$$

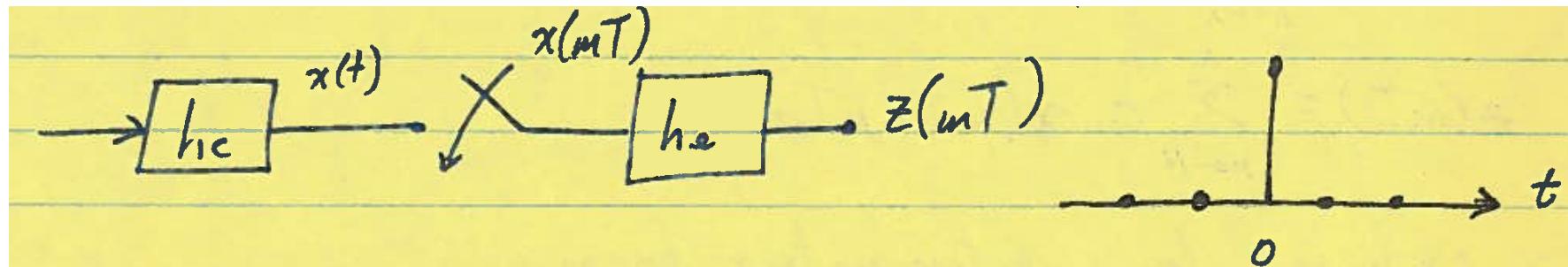


# ZFE Filter Structure

- This channel inversion can be achieved with...
  - FIR filter structure (transversal filter)

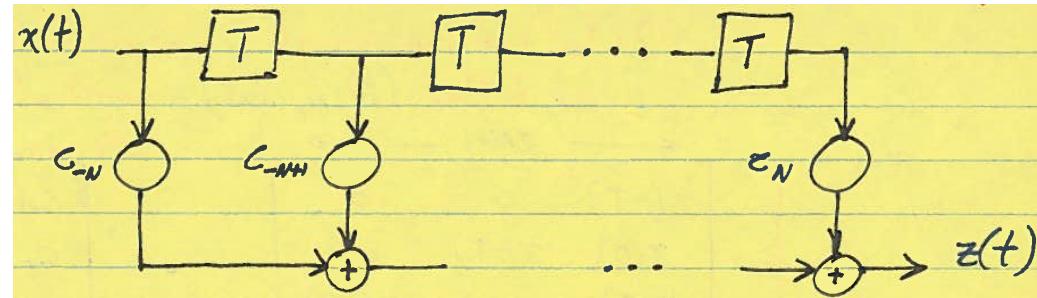


- $x$  is the equalizer input (with ISI)
- $z$  is the equalizer output (without ISI)



# ZFE: Formal Signal Description

- Math for FIR filter input/output relation...



$$z(t) = \sum_{n=-N}^N c_n x(t-nT)$$

$$z(mT) = \sum_{n=-N}^N c_n x(mT-nT)$$

- Convenient to describe WHOLE ZFE impulse response
  - i.e. the whole of  $z$  for any particular channel input

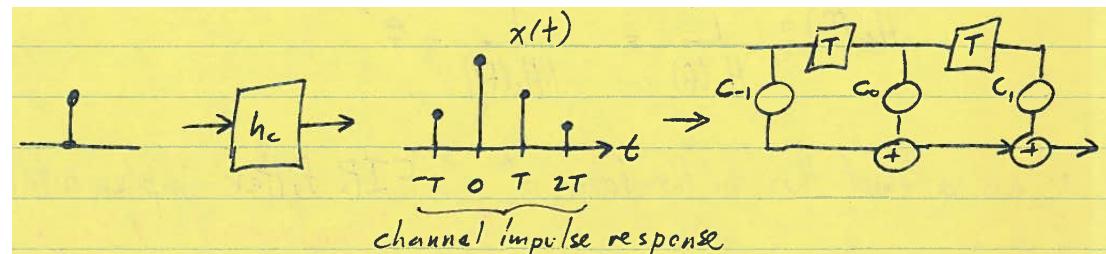
# ZFE Channel Response

- Capture entire ZFE response
  - using convolution matrix  $\mathbf{X}$

$$\bar{\mathbf{z}} = \tilde{\mathbf{X}} \cdot \bar{\mathbf{c}}$$

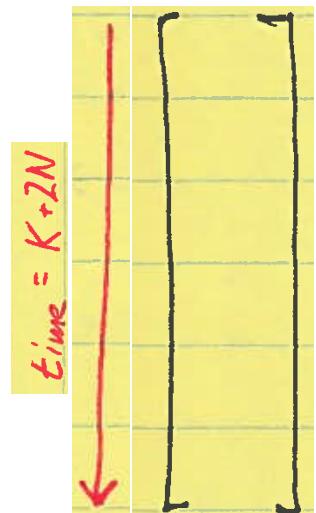
- $\mathbf{X}$ 
  - state of equalizer as a function of time
    - state: the signal value at each tap

- For...
  - a channel impulse response of **K** impulses
  - into an equalizer of **N+1** taps
- We have a **z** how long?



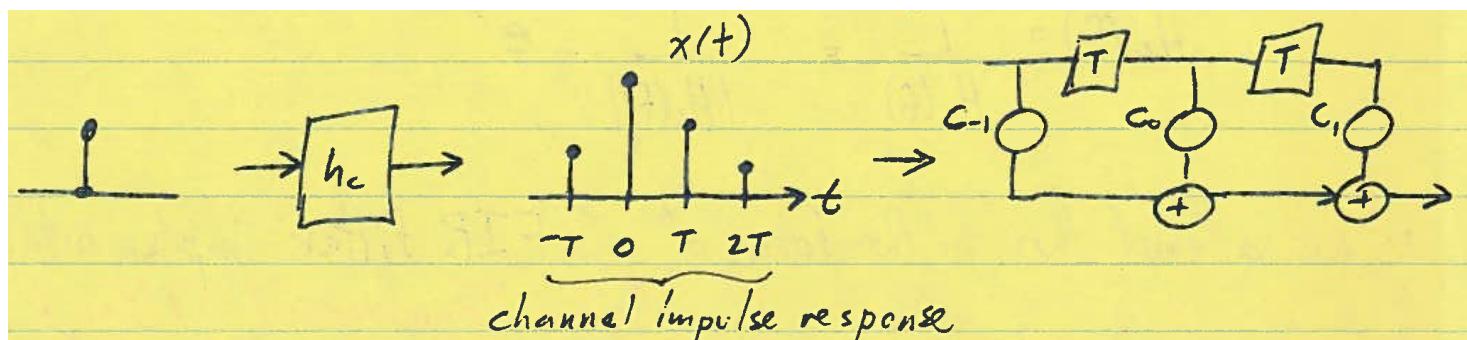
# ZFE Channel Response

$$\mathbf{N} = \bar{X} \cdot \bar{C}$$



$$= \begin{bmatrix} x(-T) & 0 & 0 \\ x(0) & x(-T) & 0 \\ x(T) & x(0) & x(-T) \\ x(2T) & x(T) & x(0) \\ 0 & x(2T) & x(T) \\ 0 & 0 & x(2T) \end{bmatrix} \cdot \begin{bmatrix} c_{-1} \\ c_0 \\ c_1 \end{bmatrix}$$

$\xleftarrow{\hspace{2cm}} 2N+1 \xrightarrow{\hspace{2cm}}$



# Desired Response

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- What output do you want from equalizer?

$$\bar{z} = \bar{X} \cdot \bar{c}$$

*your desired output vector: a 1 and the rest zeros*

- And how do you get this?

– using math

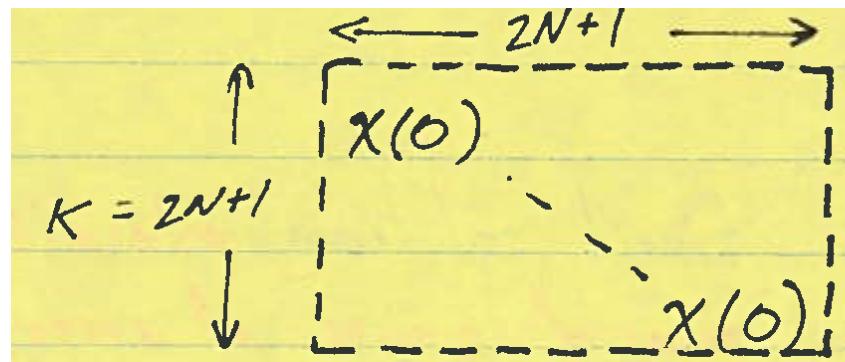
$$\bar{c} = \bar{X}^{-1} \bar{z}$$

- And what if  $X$  is not a square matrix?

# ZFE Coefficients

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- Book approach...
- ...Just make ZFE square!



# ZFE Example

- Channel response:



$$\begin{array}{c} \uparrow x(-1) \\ x(0) \\ \downarrow x(1) \\ \downarrow x(2) \\ \downarrow x(3) \end{array} \left[ \begin{array}{c} 36 \\ 230 \\ 97 \\ 37 \\ 18 \end{array} \right] mV = \bar{x}$$

- Say you want:

$$\bar{z} = \begin{bmatrix} 0 \\ -1 \\ 0 \\ 0 \\ 0 \end{bmatrix}$$

# Answer

- Square convolution matrix:

$$\bar{X} = \begin{bmatrix} 230 & 36 & 0 & 0 & 0 \\ 97 & 230 & 36 & 0 & 0 \\ 37 & 97 & 230 & 36 & 0 \\ 18 & 37 & 97 & 230 & 36 \\ 0 & 18 & 37 & 97 & 230 \end{bmatrix}$$

$$\bar{c} = \bar{X}^{-1} \bar{z} = \begin{bmatrix} -0.77 \\ 4.93 \\ -1.98 \\ 0.14 \\ -0.13 \end{bmatrix}$$

normalize using  $c_i$

$$\begin{bmatrix} -.097 \\ .624 \\ -.25 \\ .017 \\ .016 \end{bmatrix}$$

$\sum c_i$

## ZFE-LS

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- Least-squares (LS)
  - A standard approach to the solution of over determined systems

$$\bar{c}_{LS} = \arg_{\bar{c}} \min \left\| \bar{z} - \bar{X} \bar{c} \right\|^2$$

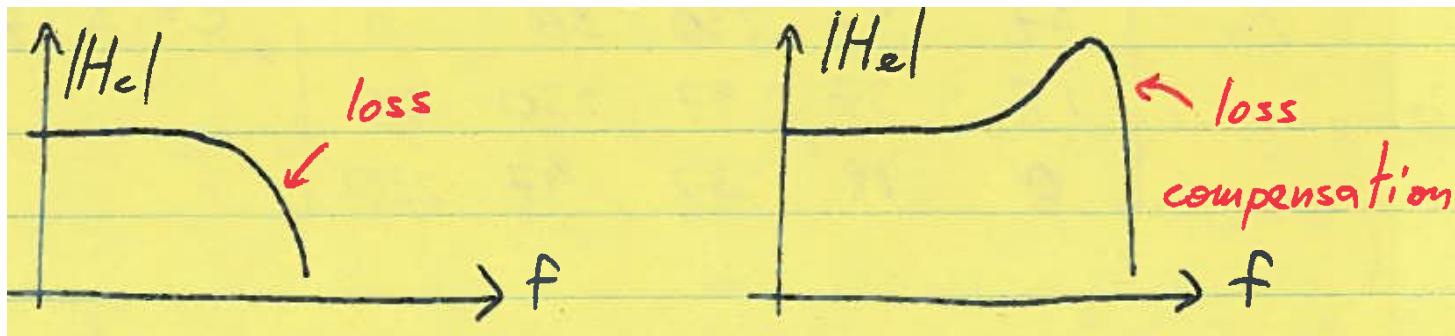
$$\sum_i (z_i - \sum_j x_{ij} c_j)^2$$

- minimize difference-squared between z and product of chosen c with x

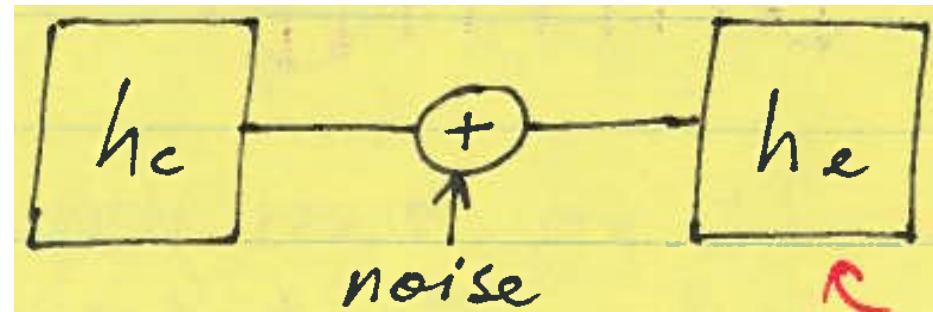
$$\bar{c}_{LS} = (\bar{X}^T \cdot \bar{X})^{-1} \bar{X}^T \cdot \bar{z}$$

# Noise Enhancement

- ZFE and ZFE-LS is SINGLE-MINDED!
- In compensating only to remove ISI...



- ...noise is amplified
  - noise enhancement
- You compensate to get back your original signal
  - ...and in the process you increase the net noise into detector



## 12.5 Minimum-Mean-Squared Equalizer

- Seek a compromise between ISI and noise enhancement
- Instead of least-squares to minimize ISI...

$$\bar{C}_{LS} = \bar{C}_{ZFE} = (\bar{X}^T \bar{X})^{-1} \bar{X}^T \bar{z}$$

- ...account for SNR

$$\bar{C}_{MMSE} = \left( \frac{\underline{I}}{SNR} + \bar{X}^T \bar{X} \right)^{-1} \bar{X}^T \bar{z}$$

identity matrix

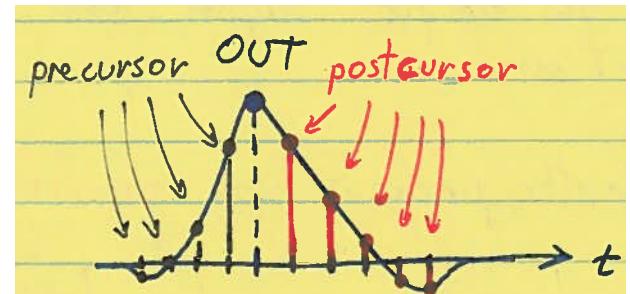
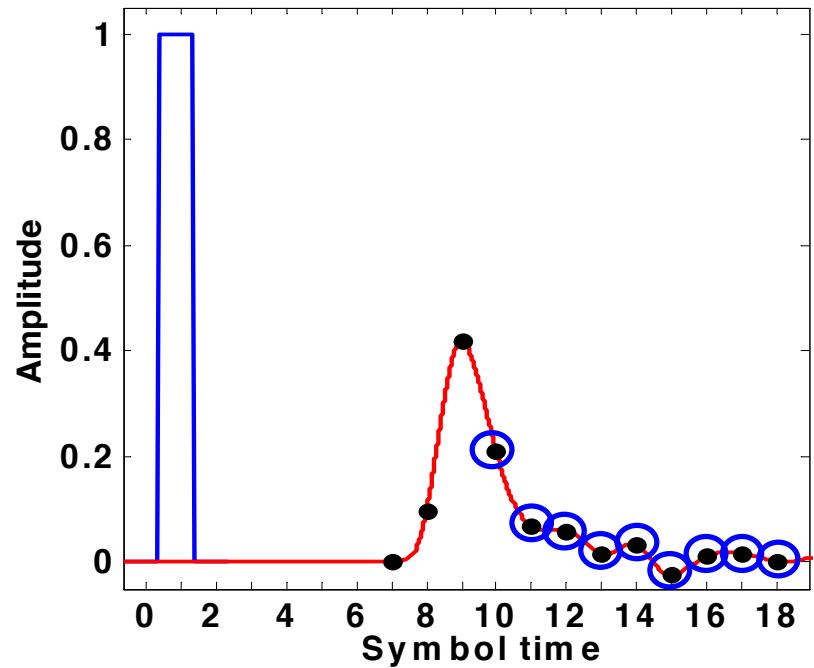
# MMSE

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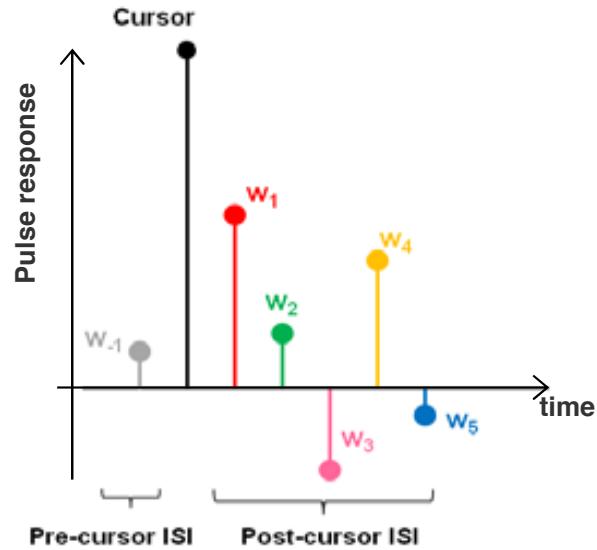
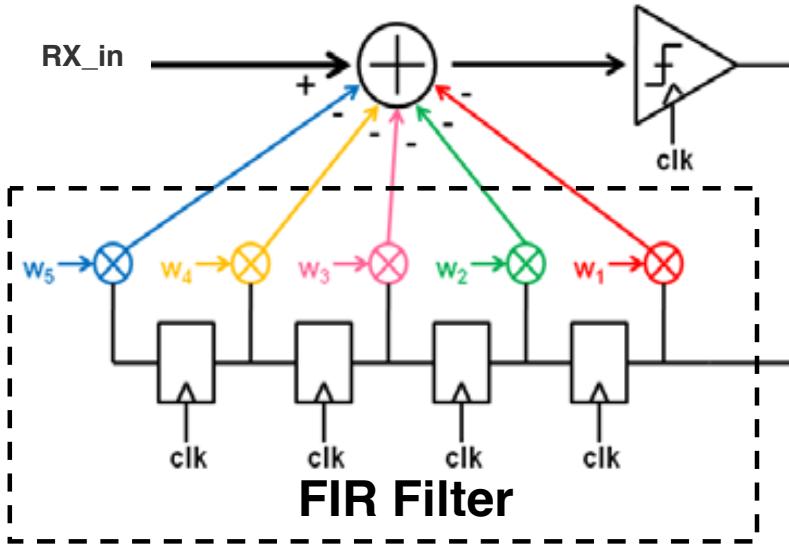
- MMSE allows a bit of ISI
  - but has less noise enhancement
- MMSE not straightforward to apply
  - noise may not be known
  - harder to make adaptive

## 12.6 Decision-Feedback Equalizer (DFE)

- A non-linear approach
- If you know which bit is transmitted you know exactly what ISI it will cause (in postcursor)
- Why not just directly cancel the coming ISI?



# DFE



- Key advantage: no noise enhancement

# DFE

- Only handles postcursor
  - May need linear filter for remove precursor

