

L14: Bandpass Modulation



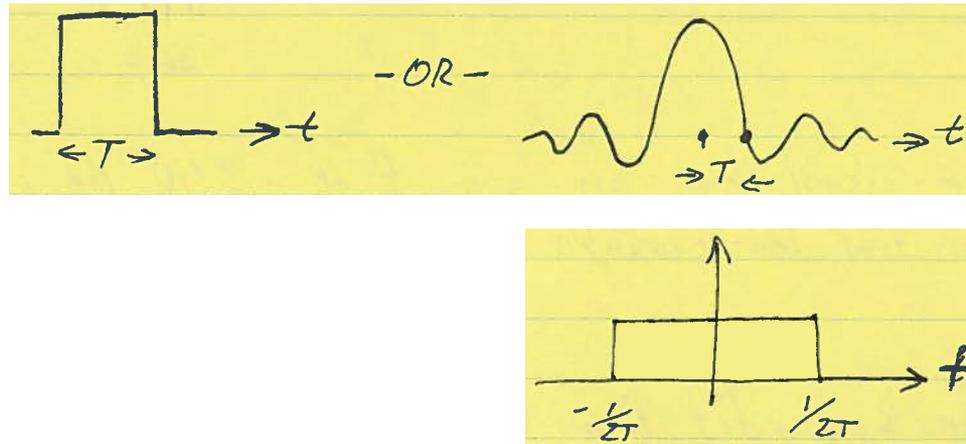
Sebastian Magierowski
York University

Outline

- 14.1 Basics
- 14.2 Bandpass Signaling
- 14.3 PSK: Phase Shift Keying
- 14.4 FSK: Frequency Shift Keying
- 14.5 ASK: Amplitude Shift Keying
- 14.6 APK: Amplitude Phase Keying
- 14.7 Signal Space Concepts
- 14.8 Signal Space Examples

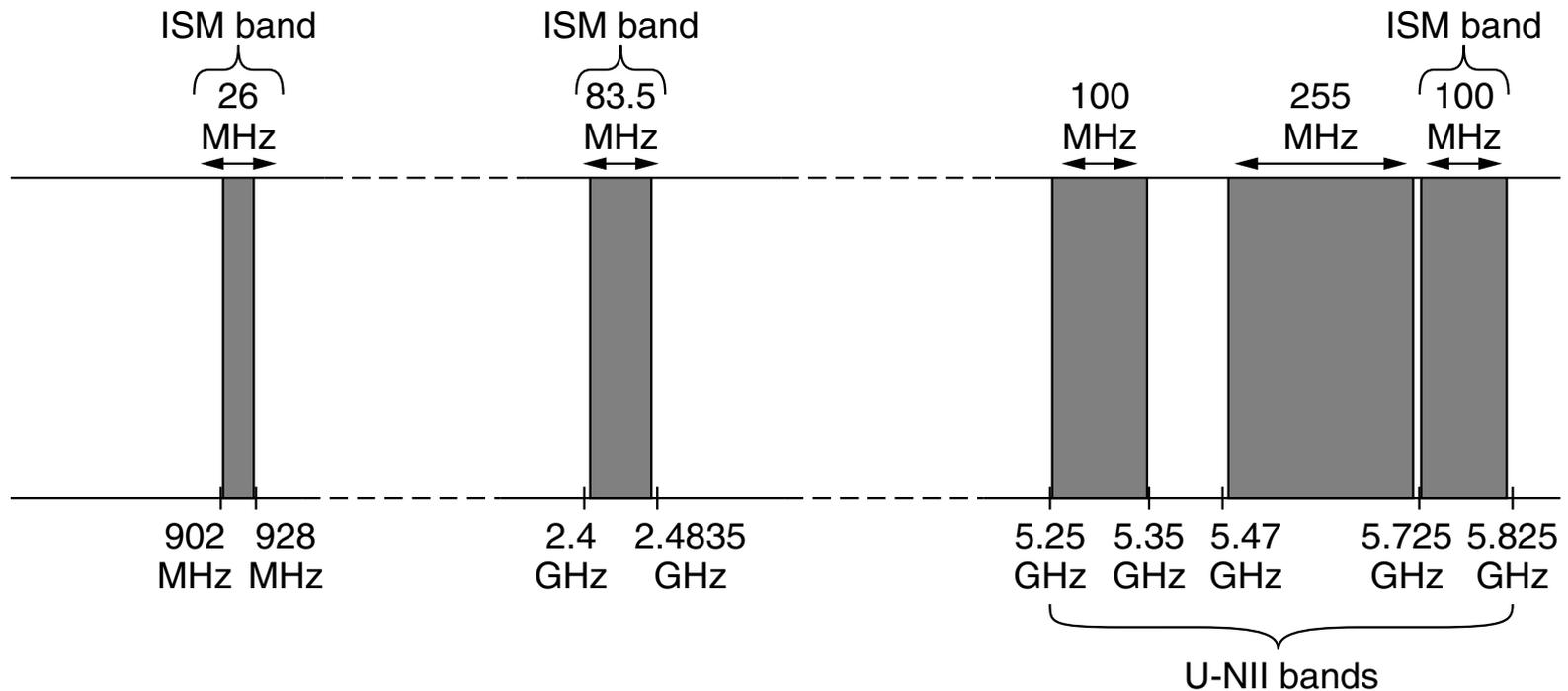
Basics: Baseband

- So far we have discussed communication pulses with spectra centred at DC
 - Basband signalling



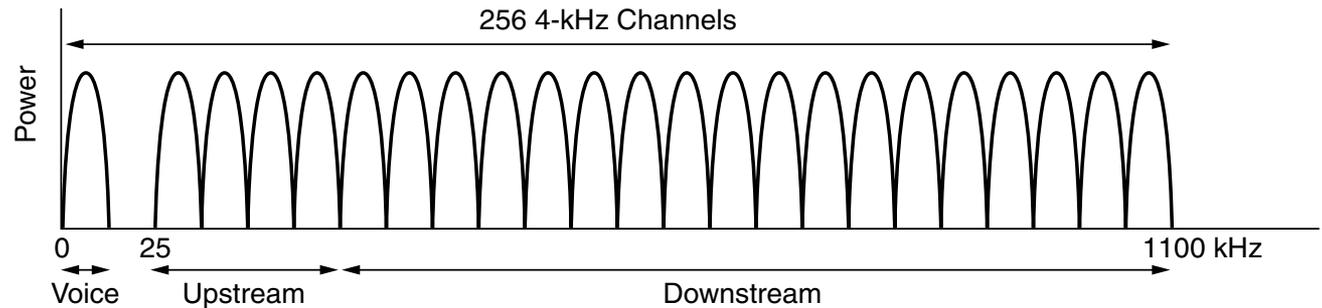
Bandpass Channels

- But there are channels not conducive to this
 - Wireless

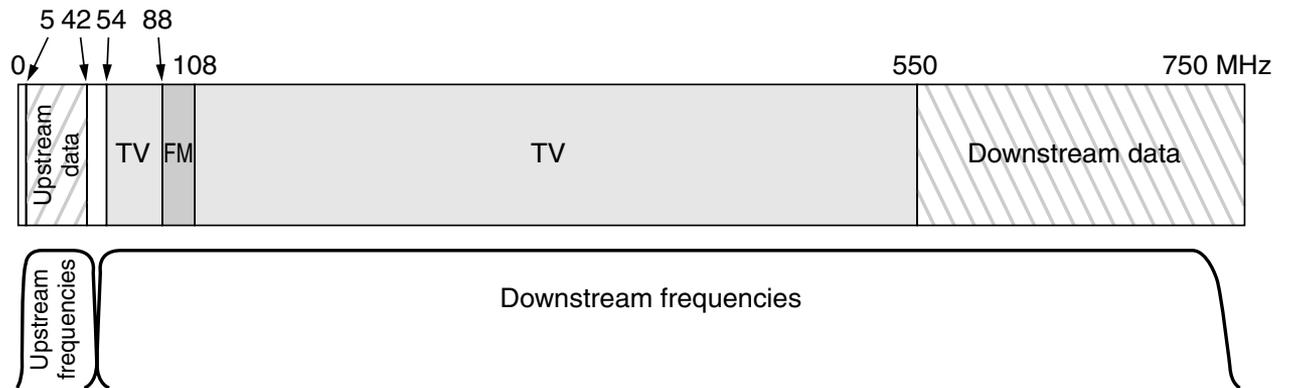


Bandpass Channels

- Wired can also be bandpass
 - ADSL

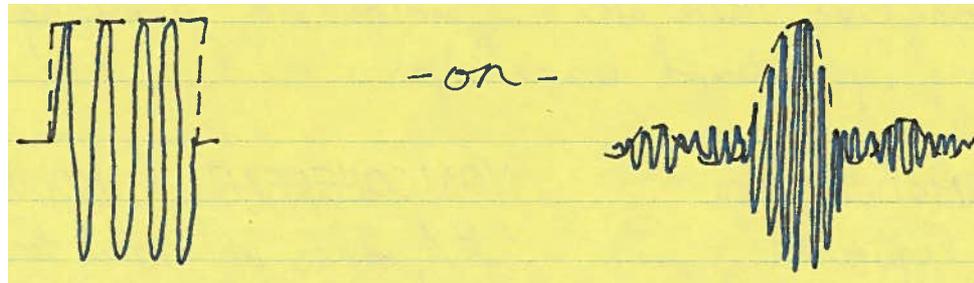
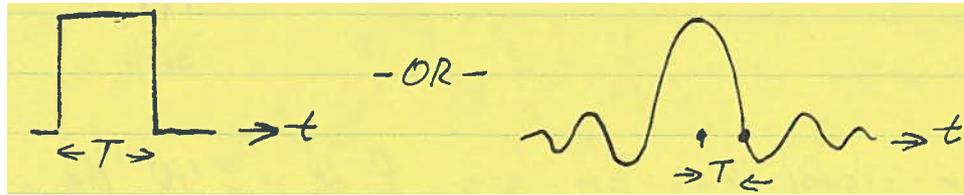


- Cable



Bandpass Signalling

- Instead of pulse you send modulated sine waves



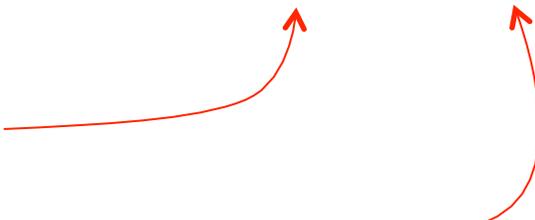
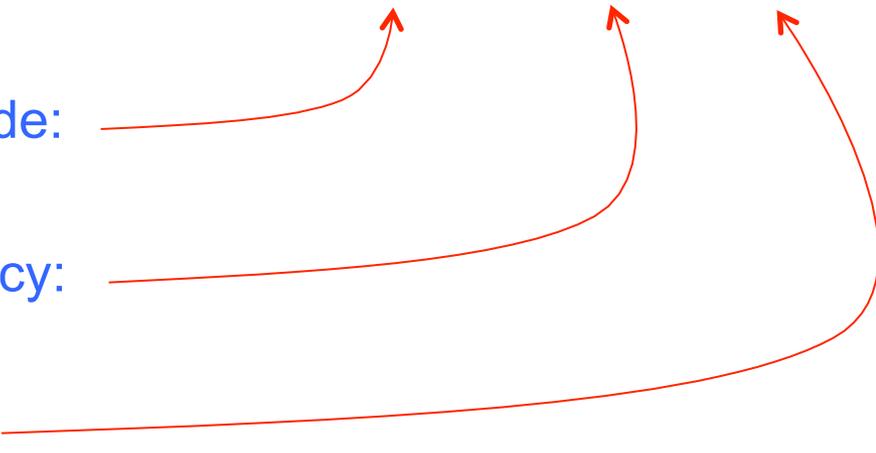
$$s(t) = A(t) \cos[\omega_0 t + \phi(t)]$$

- Which result in bandpass channels (in frequency)
 - With less spectral efficiency!
 - At first look

Bandpass Signal Parameters

- Three main parameters to control

$$s(t) = A(t) \cos[\omega_o t + \phi(t)]$$

- amplitude: 
- frequency: 
- phase: 

- These extra “levers” allow you to squeeze in more data per symbol
 - We’ll see this more clearly soon

A Note on Signal Settings

- Common to designate amplitude in terms of signal energy per symbol

$$A_{rms} = \left\{ \frac{1}{T} \int_0^T A^2 \cos^2(\omega_o t) dt \right\}^{\frac{1}{2}} = \frac{A}{\sqrt{2}}$$

$$A = \sqrt{2}A_{rms} = \sqrt{2A_{rms}^2} = \sqrt{2P} = \sqrt{\frac{2E}{T}}$$

– P = avg. power per symbol

A Peak at Demodulation

- Coherent Demodulation
 - RX has to know the phase of your sine wave to pick out the data
 - Carrier recover needed
- Non-Coherent Demodulation
 - RX does not need to know the phase of sine wave

14.3 PSK: Phase Shift Keying

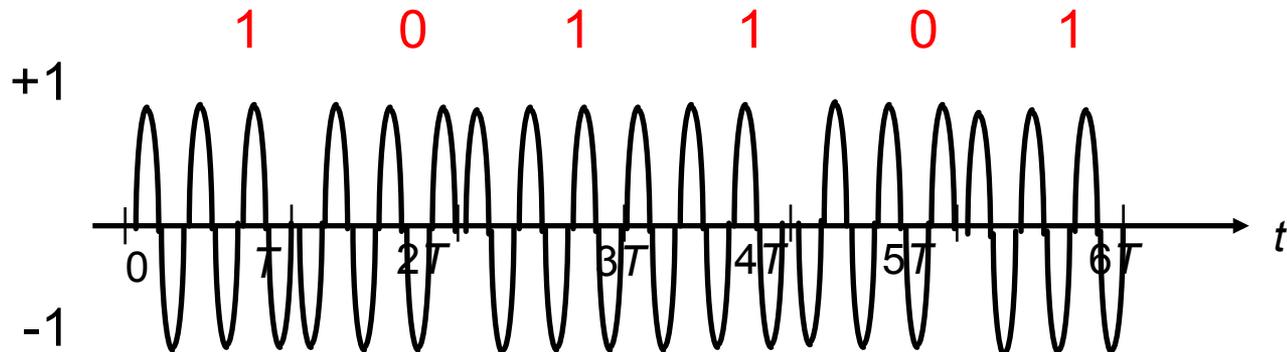
- Symbols have different phase

$$s_i(t) = \sqrt{\frac{2E}{T}} \cos[\omega_o t + \phi_i(t)], \quad 0 \leq t \leq T, i = 1, \dots, M$$

↓

$$\phi_i(t) = \frac{2\pi i}{M}$$

- BPSK: Binary Phase Shift Keying $\phi_1(t) = \pi$ $\phi_2(t) = 0$



Binary Phase
Shift Keying
(BPSK)

14.4 FSK: Frequency Shift Keying

- Symbols have different frequency
 - ω_i has M discrete values

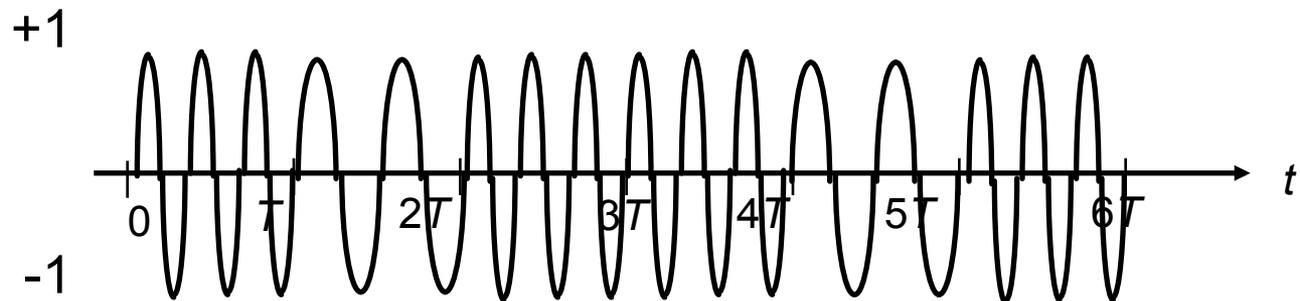
$$s_i(t) = \sqrt{\frac{2E}{T}} \cos[\omega_i t + \phi], \quad 0 \leq t \leq T$$

- Binary example:

Information

1 0 1 1 0 1

Frequency
Shift
Keying



14.5 ASK: Amplitude Shift Keying

- Amplitude takes on different value for each symbol

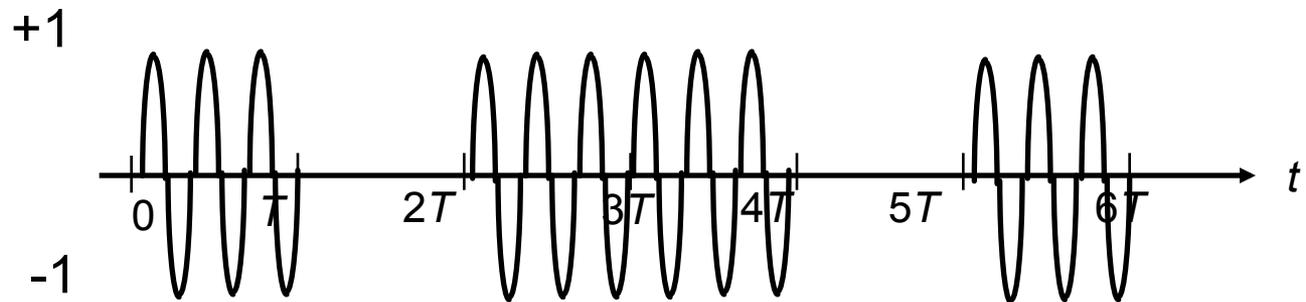
$$s_i(t) = \sqrt{\frac{2E_i}{T}} \cos[\omega_o t + \phi], \quad 0 \leq t \leq T$$

- Binary example
 - OOK (On-Off Keying)

Information

1 0 1 1 0 1

Amplitude
Shift
Keying



14.6 APK: Amplitude-Phase Keying

- A combination of ASK & PSK

$$s_i(t) = \sqrt{\frac{2E_i}{T}} \cos[\omega_o t + \phi_i], \quad 0 \leq t \leq T$$

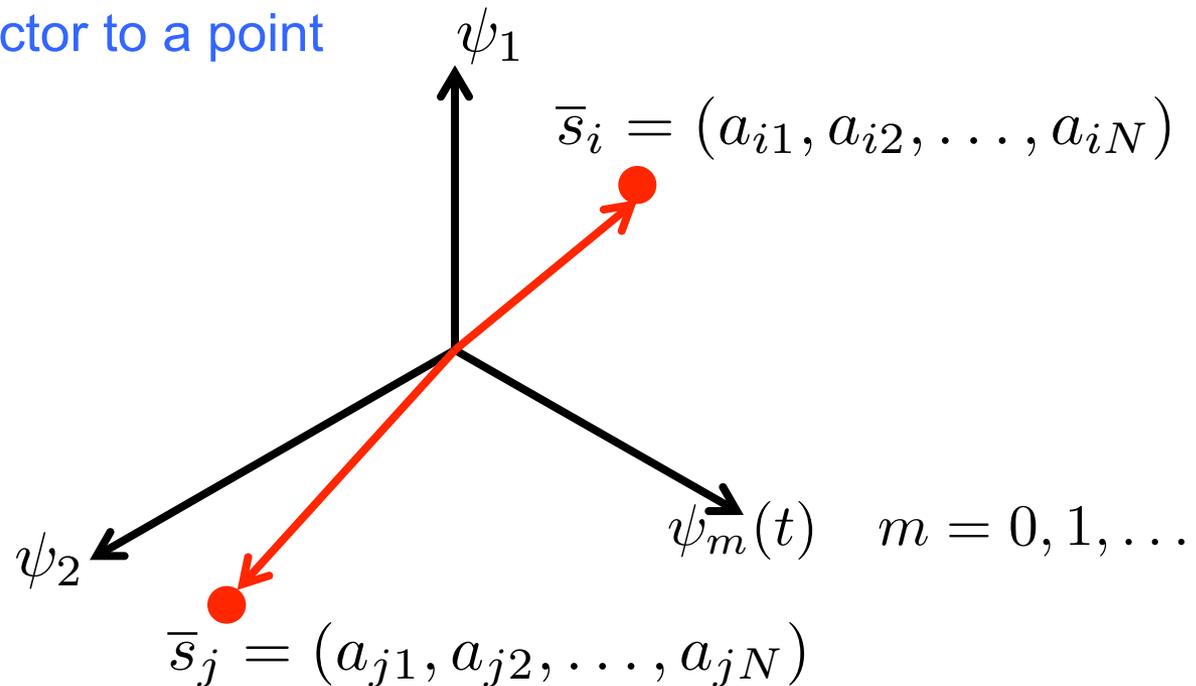
- QAM: Quadrature Amplitude Modulation

14.7 Signal-Space Concepts

- A more general description of signals is possible
 - Not just bandpass signals
 - But also signal sequences!
- The description consists of...
 - ...sequence of orthogonal signals
- These descriptions allow more efficient organization of receiver structures for sophisticated signals

Signal Space

- Your signal can be represented as a point in an abstract space
 - Rather a vector to a point



- The axes represent orthogonal signals
 - Which constitute any signal you may wish to define in this space

Basis Functions

- Basis Functions

- The functions that define our signal space coordinates

$$\psi_m(t) \quad m = 0, 1, \dots$$

- Defined over some time interval T_i to T_f
 - T_i : initial time, T_f : final time
- Make sure the functions are PERPENDICULAR to each other
 - Orthonormal

$$\int_{T_i}^{T_f} \psi_i(t)\psi_j(t)dt = \delta_{ij} = \begin{cases} 0, & \text{if } i \neq j \\ 1, & \text{if } i = j \end{cases}$$

Basis Function Example

- Fourier series coefficients

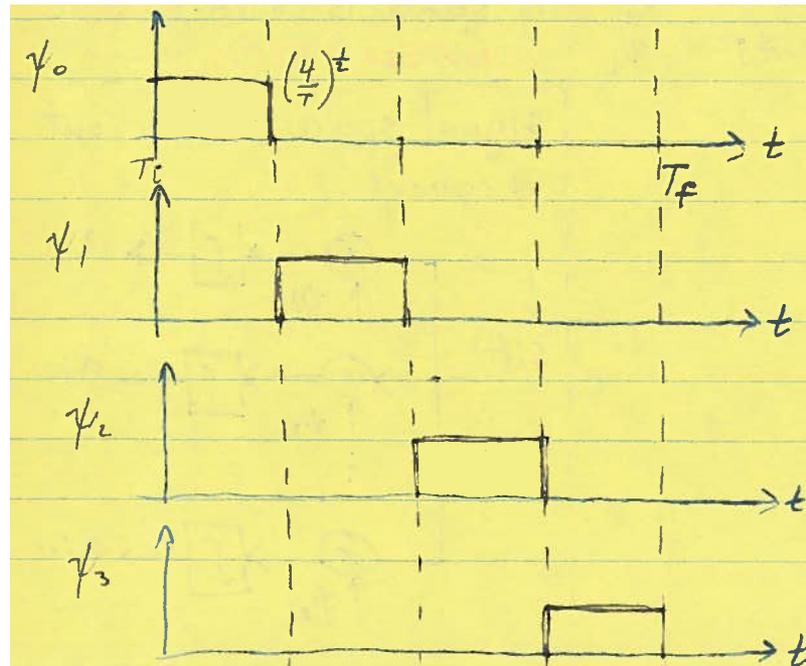
$$\psi_m(t) = \left\{ \left(\frac{1}{T}\right)^{\frac{1}{2}}, \left(\frac{2}{T}\right)^{\frac{1}{2}} \cos(m\omega_o t), \left(\frac{2}{T}\right)^{\frac{1}{2}} \sin(m\omega_o t), \dots \right\} \quad m = 1, 2, \dots$$

$$T = T_f - T_i$$

$$T = \frac{2\pi}{\omega_o}$$

Basis Function Example

- Finite set of N non-overlapping pulses



Signals in Space

- Any of M symbols: $s_i(t)$, $i = 1, 2, \dots, M$
- Can be expressed as a N -term series expansion
 - of orthonormal basis functions

$$s_i(t) = \sum_{j=1}^N a_{ij} \psi_j(t)$$

- expansion coefficients a_{ij}

$$a_{ij} = \int_{T_i}^{T_f} s_i(t) \psi_j(t) dt$$

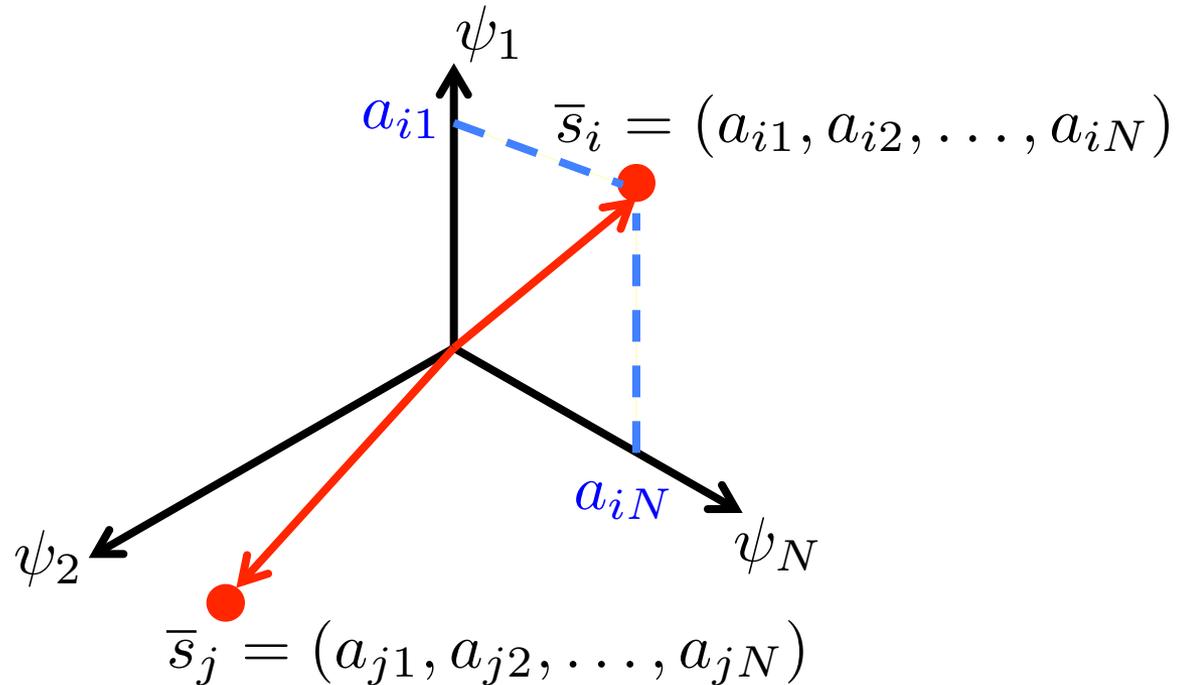
- projection of i th symbol on j th basis function

Signal Space Pictorially

- Collection of M points in N-space
 - Signal Constellation

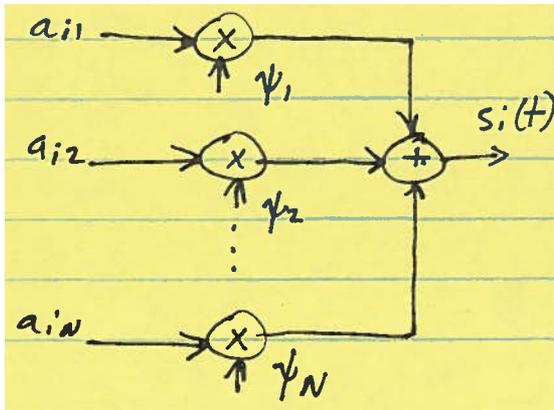
$$s_i(t) = \sum_{j=1}^N a_{ij} \psi_j(t)$$

$$a_{ij} = \int_{T_i}^{T_f} s_i(t) \psi_j(t) dt$$



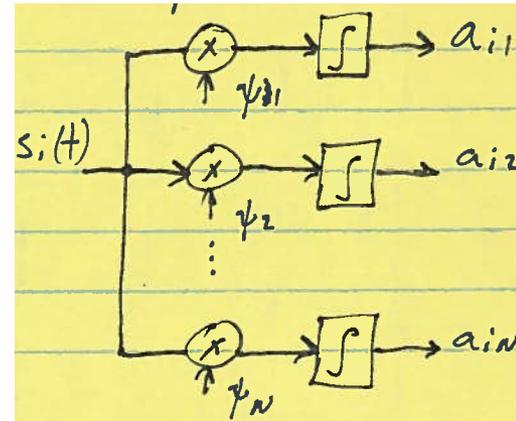
Signal Generation & Recovery

- Generation



$$s_i(t) = \sum_{j=1}^N a_{ij} \psi_j(t)$$

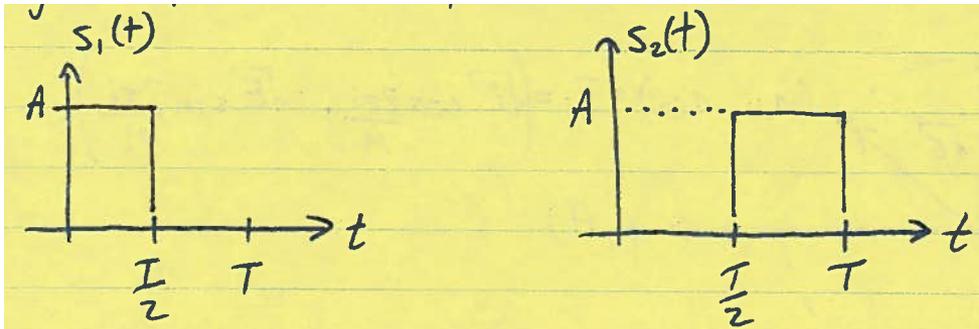
- Recovery



$$a_{ij} = \int_{T_i}^{T_f} s_i(t) \psi_j(t) dt$$

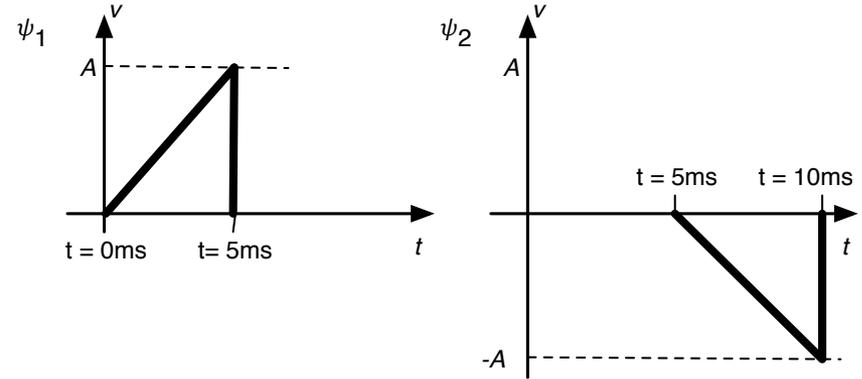
14.8 Signal Space Examples

- Draw the signal constellation



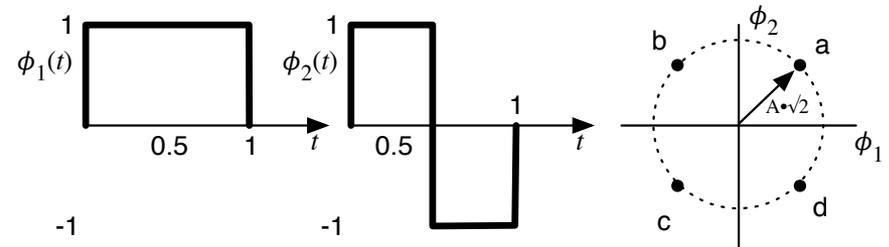
Signal Space Example

- What value of A is needed to make these basis functions orthonormal?



Signal Space Example

- For the basis functions shown, sketch the signal waveforms corresponding to the indicated constellation points



M-ary PSK in Signal Space

- The M-ary PSK

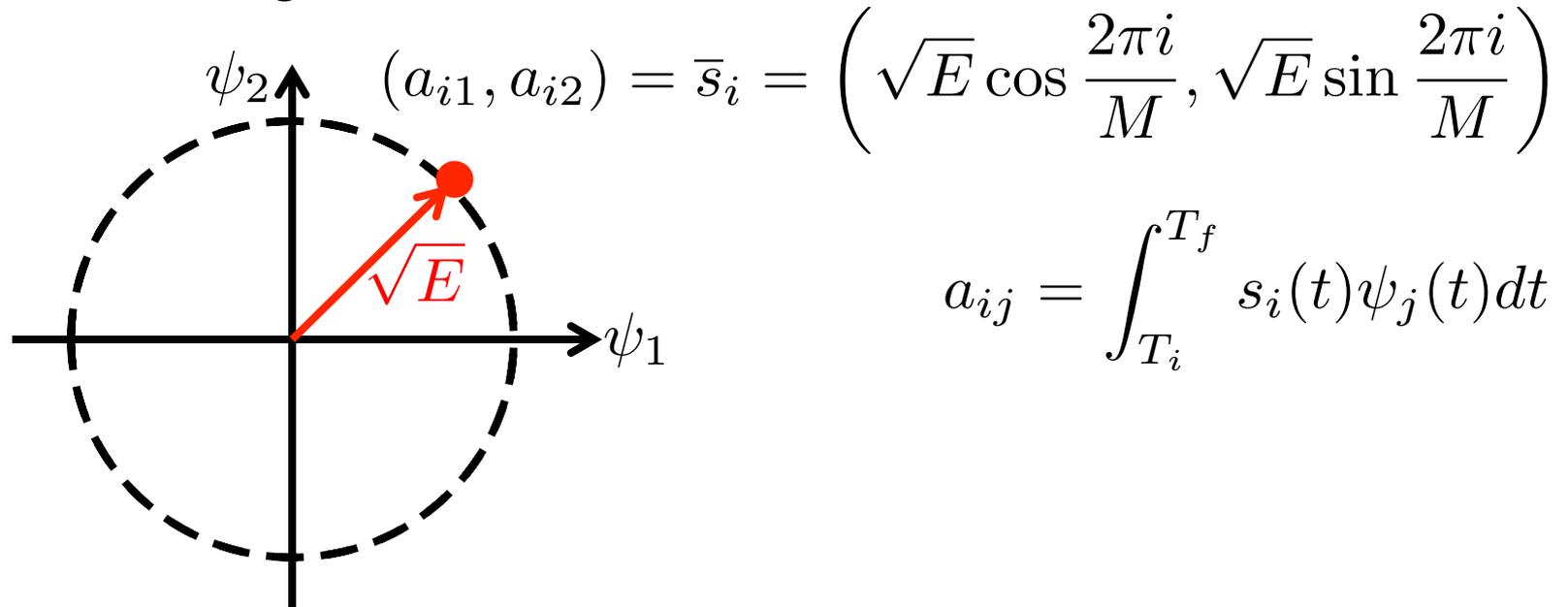
$$s_i(t) = \sqrt{\frac{2E}{T}} \cos\left(\omega_o t + \frac{2\pi i}{M}\right), \quad i = 0, \dots, M - 1$$

- Just a harmonic signal with some phase
- Fourier series only needs 2 coefficients for this
 - Hence our basis functions can be...

$$\psi_1(t) = \sqrt{\frac{2}{T}} \cos(\omega_o t) \qquad \psi_2(t) = \sqrt{\frac{2}{T}} \sin(\omega_o t)$$

M-ary PSK Constellation

- The resulting constellation is



$$a_{ij} = \int_{T_i}^{T_f} s_i(t) \psi_j(t) dt$$

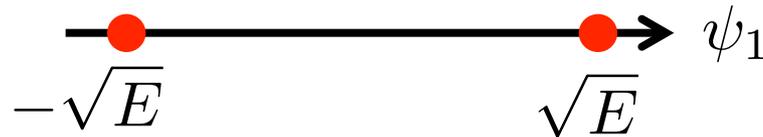
$$s_i(t) = \sqrt{\frac{2E}{T}} \cos \left(\omega_o t + \frac{2\pi i}{M} \right), \quad i = 0, \dots, M - 1$$

2-ary PSK = BPSK

- Constellation is... $\phi_1(t) = 0$ $\phi_2(t) = \pi$

$$-\sqrt{\frac{2E}{T}} \cos(\omega_o t) = s_2(t)$$

$$s_1(t) = \sqrt{\frac{2E}{T}} \cos(\omega_o t)$$



$$\psi_1(t) = \sqrt{\frac{2}{T}} \cos(\omega_o t)$$

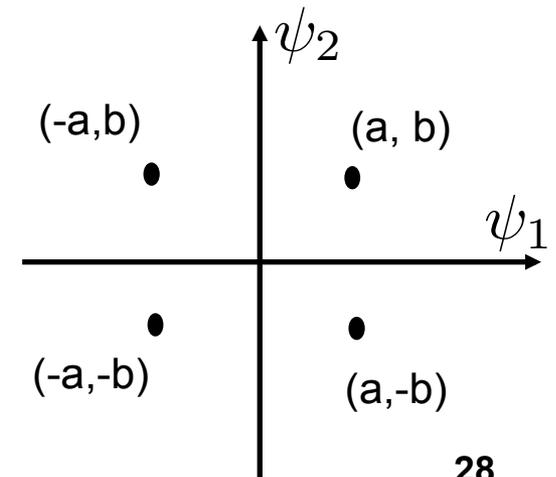
QAM

- Quadrature Amplitude Modulation

$$s_i(t) = a_{i1} \sqrt{\frac{2}{T}} \cos(\omega_o t) + a_{i2} \sqrt{\frac{2}{T}} \sin(\omega_o t)$$

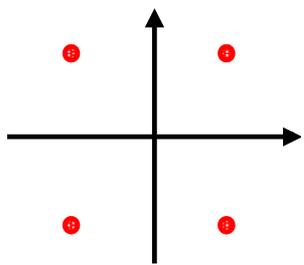
$$s_i(t) = a_i \sqrt{\frac{2}{T}} \cos(\omega_o t) + b_i \sqrt{\frac{2}{T}} \sin(\omega_o t)$$

- Signal space coordinates: (a_i, b_i)
 - Typically chosen from points on a 2D square grid
 - 4-QAM example

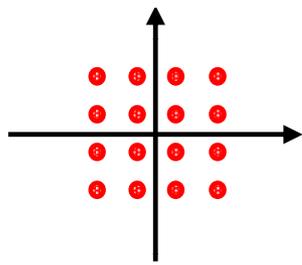


Larger QAM Constellations

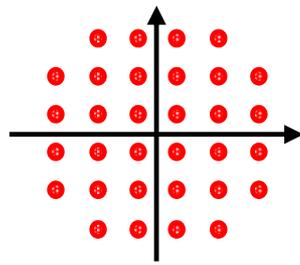
- With 4-QAM
 - bits represented per symbol: $N = 2$
 - Constellation points: $M = 2^N = 4$
- Many other possibilities (rectangular constellation)
 - $N = 3, 4, 5, \dots$
 - Even N preferred (easier coder)



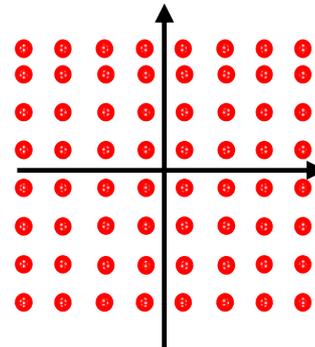
$M = 4$
 $L = 2$



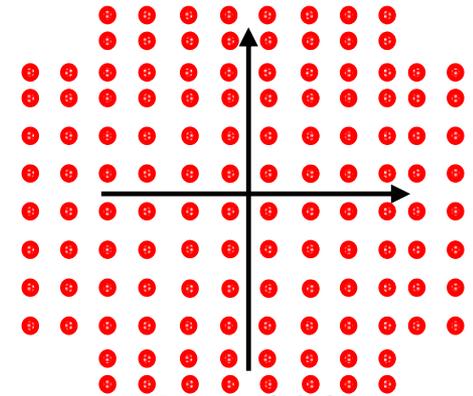
$M = 16$
 $L = 4$



$M = 32$
 $L = 6$



$M = 64$
 $L = 8$



$M = 128$
 $L = 12$

- Increasing N requires **more power**